SGM8632C
480μA, 6MHz, Rail-to-Rail I/O CMOS Operational Amplifier

GENERAL DESCRIPTION

The dual SGM8632C is a low noise, low voltage, and low power operational amplifier that can be designed into a wide range of applications. The SGM8632C has a high gain-bandwidth product of 6MHz, a slew rate of 3.7V/μs and a quiescent current of 480μA/amplifier at 5V.

The SGM8632C is designed to provide optimal performance in low voltage and low noise systems. It provides rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV. The operating supply range is from 2V to 5.5V.

The dual SGM8632C is available in Green MSOP-8 package. It is specified over the extended industrial temperature range (-40℃ to +125℃).

FEATURES

- **Rail-to-Rail Input and Output**
- **3.5mV Maximum Input Offset Voltage**
- **High Gain-Bandwidth Product: 6MHz**
- **High Slew Rate: 3.7V/μs**
- **Settling Time to 0.1% with 2V Step: 0.5μs**
- **Overload Recovery Time: 0.9μs**
- **Low Noise: 13nV/$\sqrt{Hz}$ at 1kHz**
- **Supply Voltage Range: 2V to 5.5V**
- **Input Voltage Range: -0.1V to +5.6V with V_S = 5.5V**
- **Low Supply Current: 480μA/Amplifier (TYP)**
- **Available in Green MSOP-8 Package**

APPLICATIONS

Sensors
Audio
Active Filters
A/D Converters
Communications
Test Equipment
Cellular and Cordless Phones
Laptops and PDAs
Photodiode Amplification
Battery-Powered Instrumentation

PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION
SGM8632C	MSOP-8	-40℃ to +125℃	SGM8632CXMS8G/TR	SGM8632 XMS XXXXX	Tape and Reel, 4000

NOTE: XXXXX = Date Code and Vendor Code.

Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

ABSOLUTE MAXIMUM RATINGS

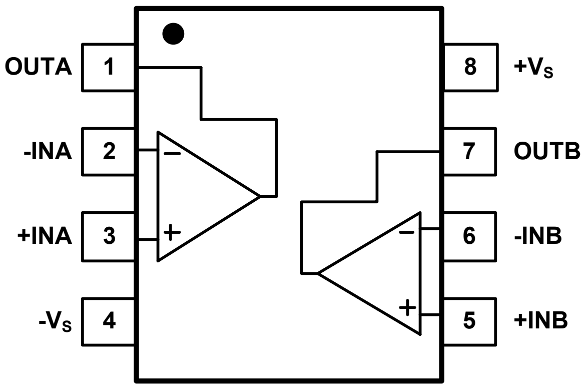
Supply Voltage, $+V_S$ to $-V_S$ 6V
Input Common Mode Voltage Range
.... $(-V_S) - 0.3V$ to $(+V_S) + 0.3V$
Package Thermal Resistance @ T_A = +25℃
MSOP-8, θ_{JA} .... 182℃/W
Storage Temperature Range ....-65℃ to +150℃
Junction Temperature...+150℃
Lead Temperature (Soldering 10sec) ....+260℃
ESD Susceptibility
HBM...8000V
MM...400V
CDM 1000V

RECOMMENDED OPERATING CONDITIONS

Operating Temperature Range ....-40℃ to +125℃

PIN CONFIGURATION

(TOP VIEW)



MSOP-8

OVERSTRESS CAUTION

Stresses beyond those listed may cause permanent damage to the device. Functional operation of the device at these or any other conditions beyond those indicated in the operational section of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

ESD SENSITIVITY CAUTION

This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

DISCLAIMER

SG Micro Corp reserves the right to make any change in circuit design, specification or other related things if necessary without notice at any time.

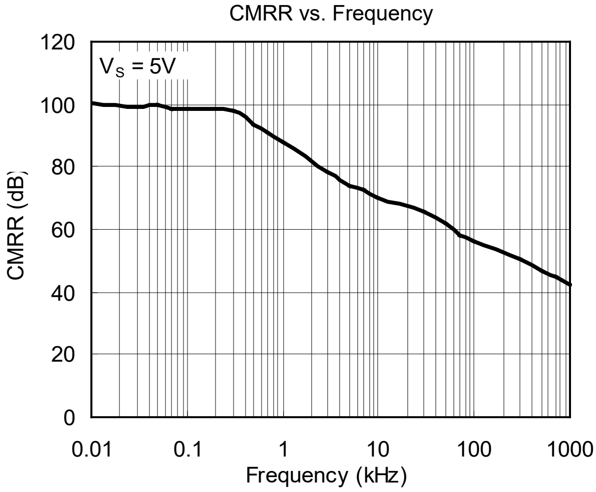
ELECTRICAL CHARACTERISTICS

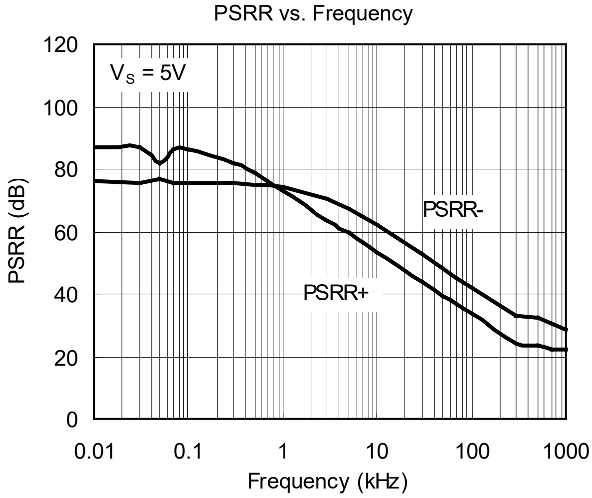
(At T_A = +25℃, V_S = 5V, V_{CM} = V_S/2, R_L = 600Ω, unless otherwise noted.)

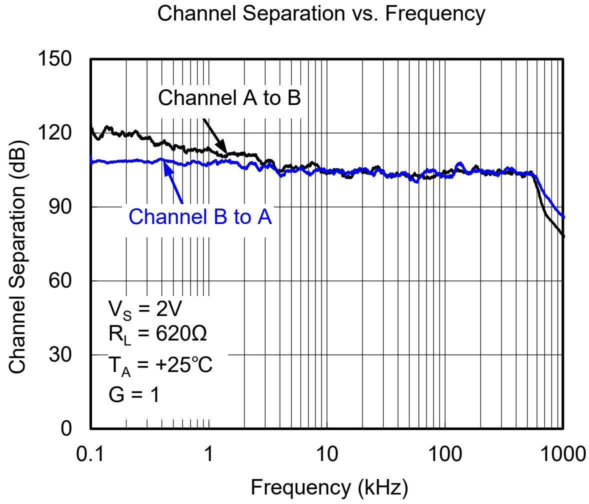
PARAMETER	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE				
		+25℃	+25℃	-40℃ to +85℃	-40℃ to +125℃	UNITS	MIN/ MAX
INPUT CHARACTERISTICS							
Input Offset Voltage (V_{OS})		0.9	3.5	3.7	3.8	mV	MAX
Input Bias Current (I_B)		1				pA	TYP
Input Offset Current (I_{OS})		1				pA	TYP
Input Common Mode Voltage Range (V_{CM})	V_S = 5.5V	-0.1 to 5.6				V	TYP
Common Mode Rejection Ratio (CMRR)	V_S = 5.5V, V_{CM} = -0.1V to 4V	84	68	67	66	dB	MIN
	V_S = 5.5V, V_{CM} = -0.1V to 5.6V	76				dB	MIN
Open-Loop Voltage Gain (A_{OL})	R_L = 600Ω, V_{OUT} = 0.15V to 4.85V	86	79	73	69	dB	MIN
	R_L = 10kΩ, V_{OUT} = 0.05V to 4.95V	103				dB	MIN
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta T$)		2.4				μV/℃	TYP
OUTPUT CHARACTERISTICS							
Output Voltage Swing from Rail	R_L = 600Ω	0.079				V	TYP
	R_L = 10kΩ	0.007				V	TYP
Output Current (I_{OUT})		58	40	30	26	mA	MIN
Closed-Loop Output Impedance	f = 200kHz, G = 1	5.4				Ω	TYP
POWER SUPPLY							
Operating Voltage Range		2	2	2	2	V	MIN
		5.5	5.5	5.5	5.5	V	MAX
Power Supply Rejection Ratio (PSRR)	V_S = 2V to 5.5V, V_{CM} = $(-V_S)$ + 0.5V	84	69	68	67	dB	MIN
Quiescent Current/Amplifier (I_Q)	I_{OUT} = 0	480	620	720	790	μA	MAX
DYNAMIC PERFORMANCE							
Gain-Bandwidth Product (GBP)		6				MHz	TYP
Phase Margin (φ_O)		63				°	TYP
Full Power Bandwidth (BW_P)	<1% distortion	250				kHz	TYP
Slew Rate (SR)	G = 1, 2V output step	3.7				V/μs	TYP
Settling Time to 0.1% (t_S)	G = 1, 2V output step	0.5				μs	TYP
Overload Recovery Time	V_{IN} × Gain = V_S	0.9				μs	TYP
NOISE PERFORMANCE							
Voltage Noise Density (e_n)	f = 1kHz	13				$nV/\sqrt{Hz}$	TYP

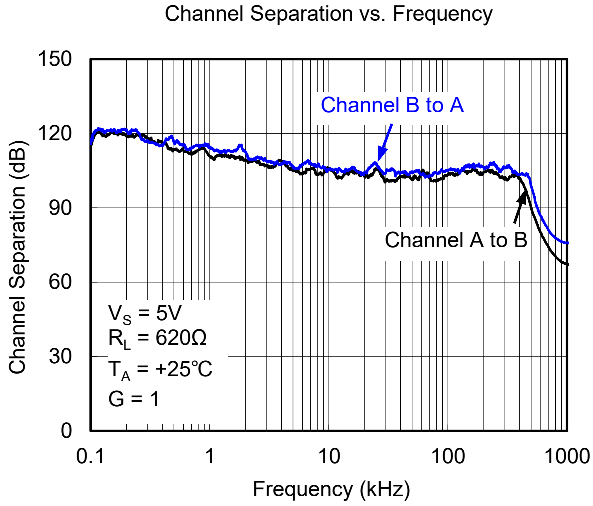
TYPICAL PERFORMANCE CHARACTERISTICS

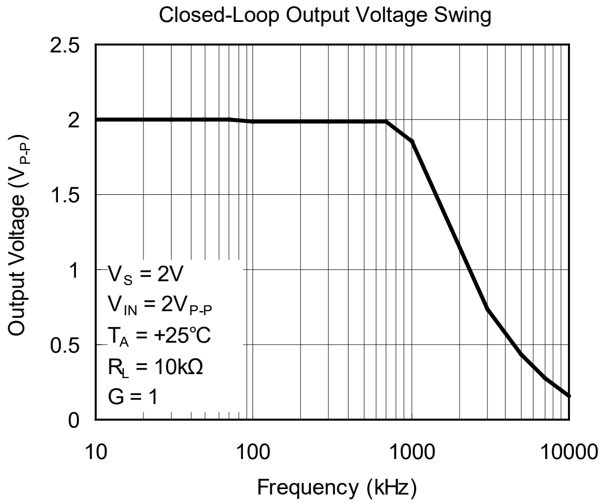
At T_A = +25°C, V_{CM} = V_S/2, R_L = 600Ω, unless otherwise noted.

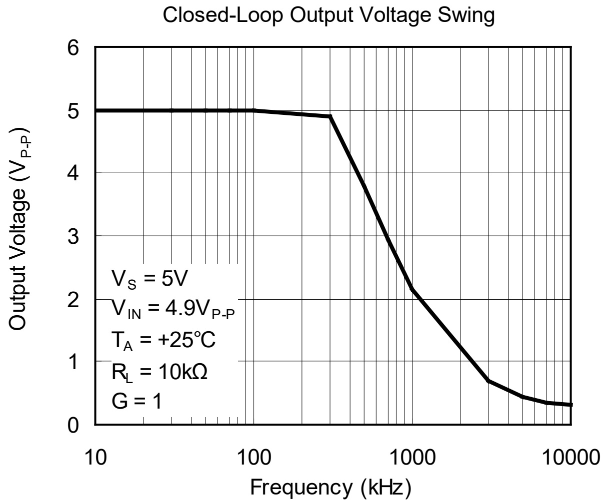






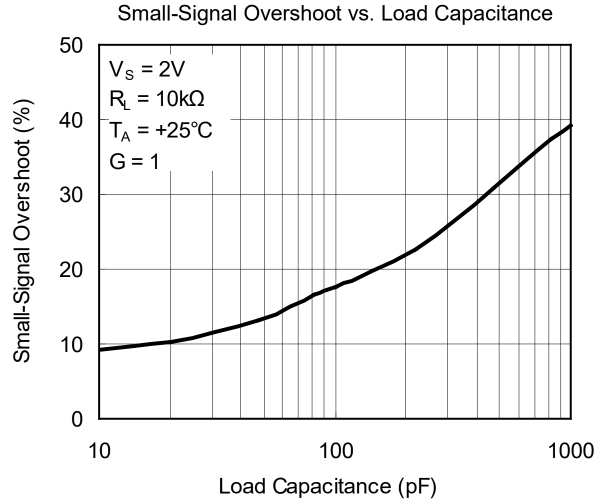


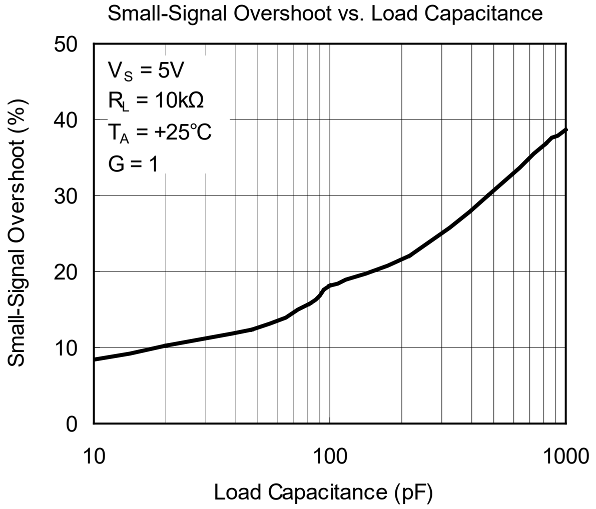


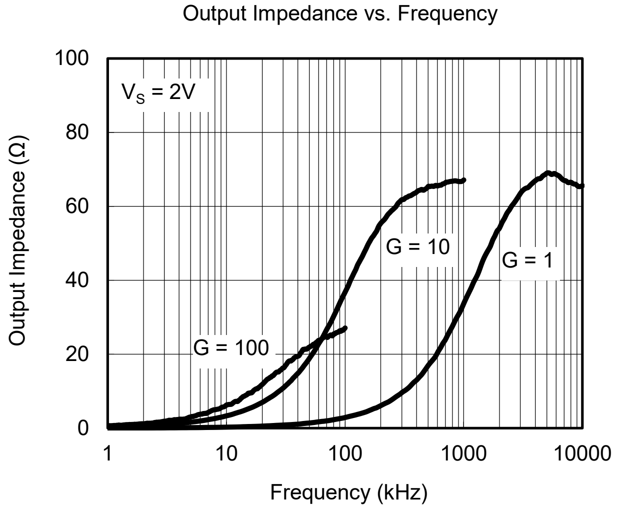


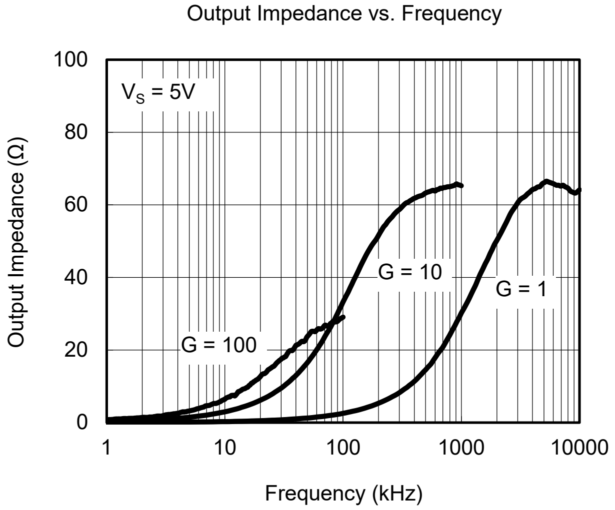
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

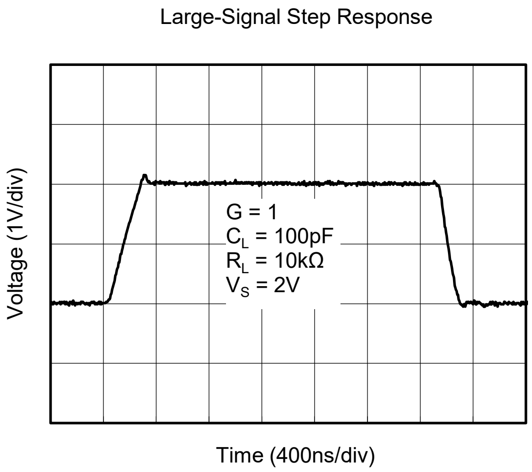
At T_A = +25℃, V_{CM} = V_S/2, R_L = 600Ω, unless otherwise noted.

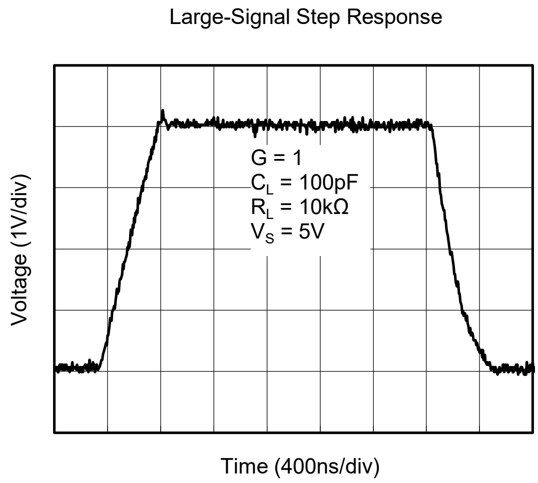








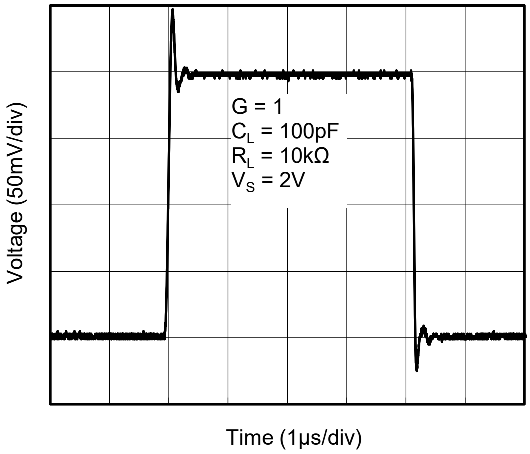




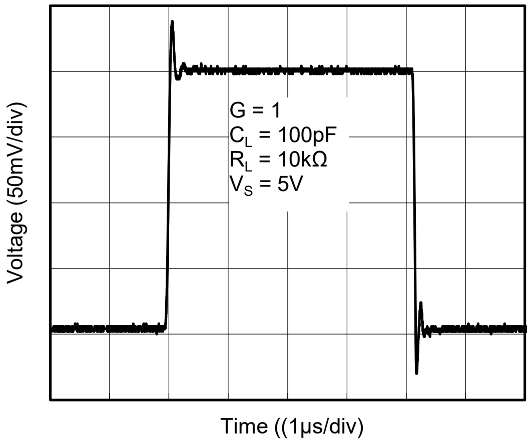
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At T_A = +25℃, V_{CM} = V_S/2, R_L = 600Ω, unless otherwise noted.

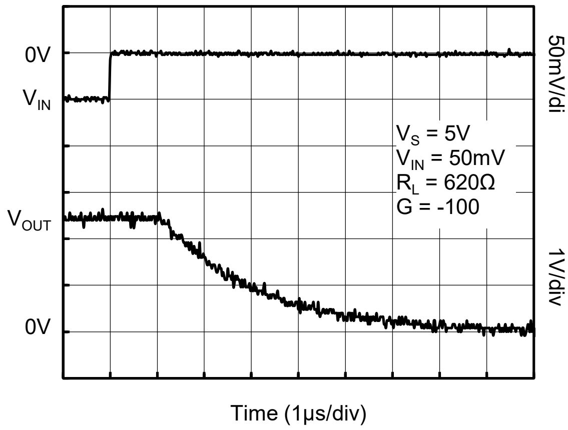
Small-Signal Step Response



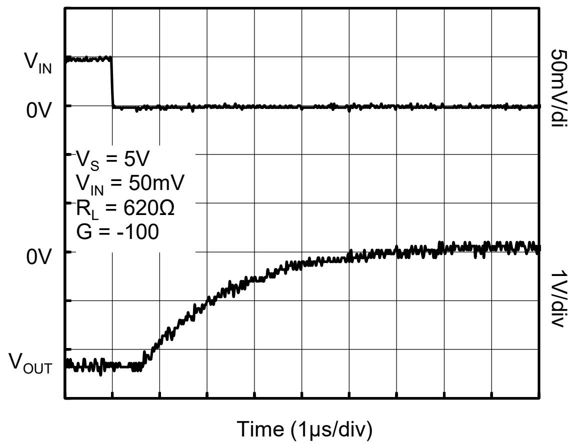
Small-Signal Step Response



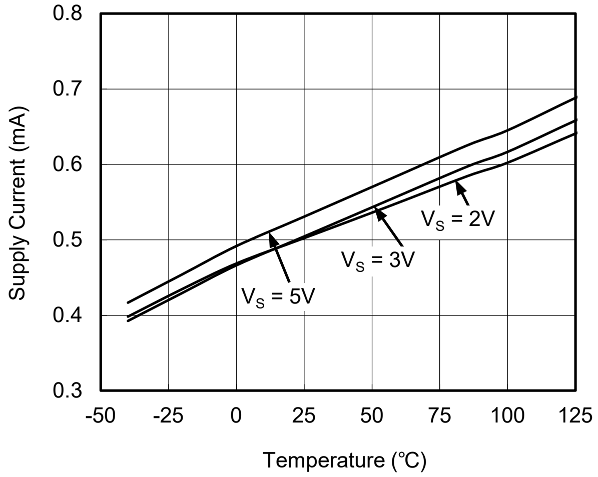
Positive overload Recovery



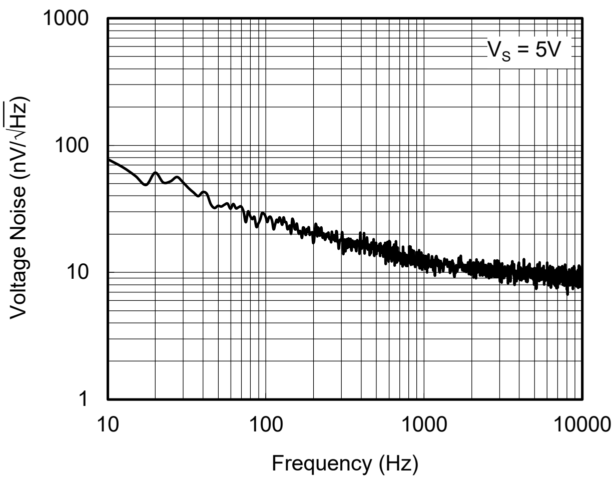
Negative Overload Recovery



Supply Current vs. Temperature



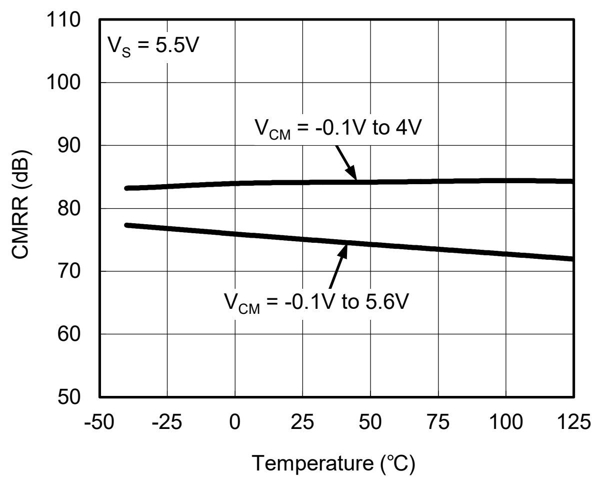
Input Voltage Noise Spectral Density vs. Frequency



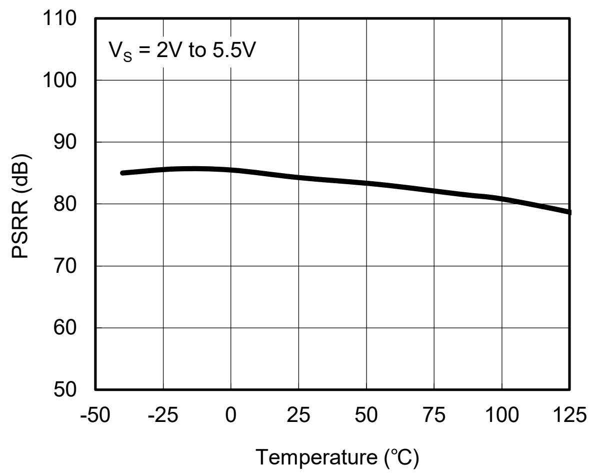
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At T_A = +25℃, V_{CM} = V_S/2, R_L = 600Ω, unless otherwise noted.

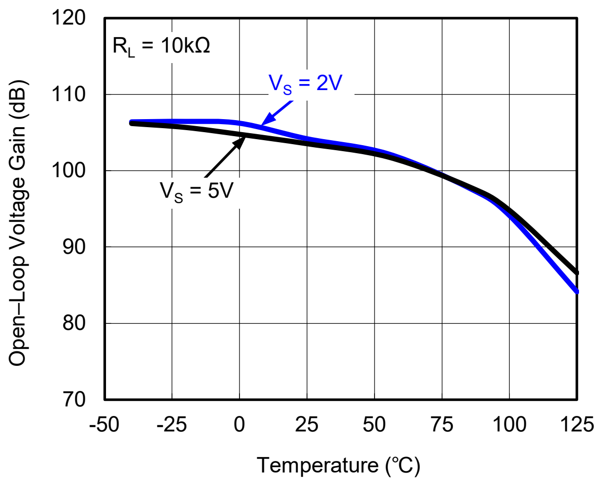
CMRR vs. Temperature



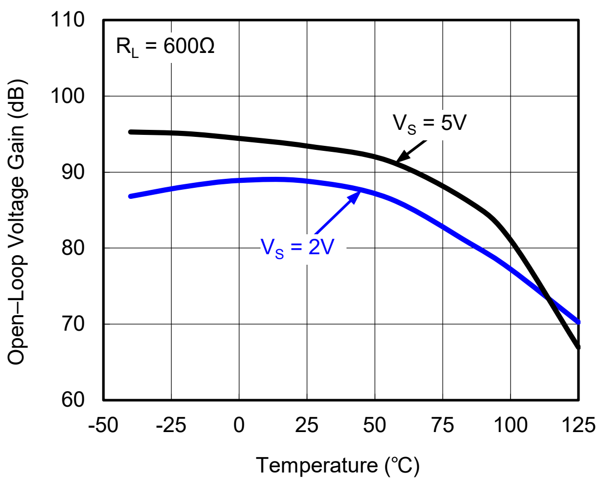
PSRR vs. Temperature



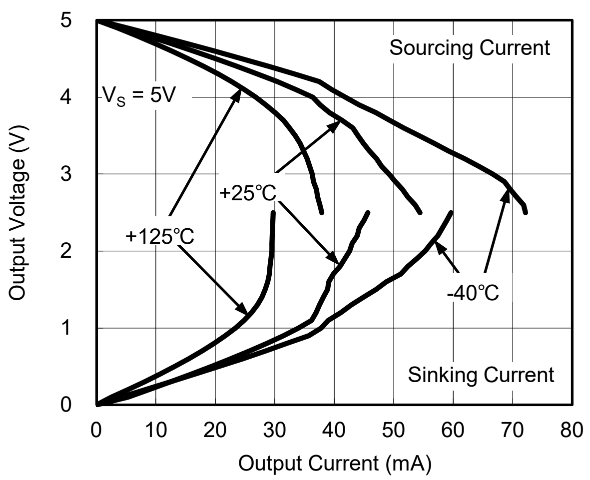
Open-Loop Voltage Gain vs. Temperature



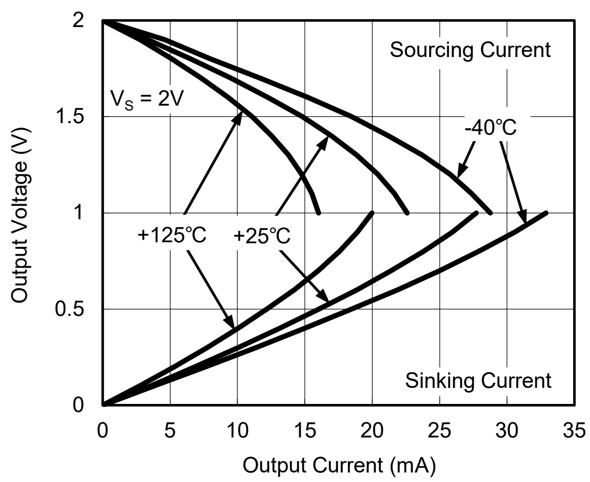
Open-Loop Voltage Gain vs. Temperature



Output Voltage Swing vs. Output Current

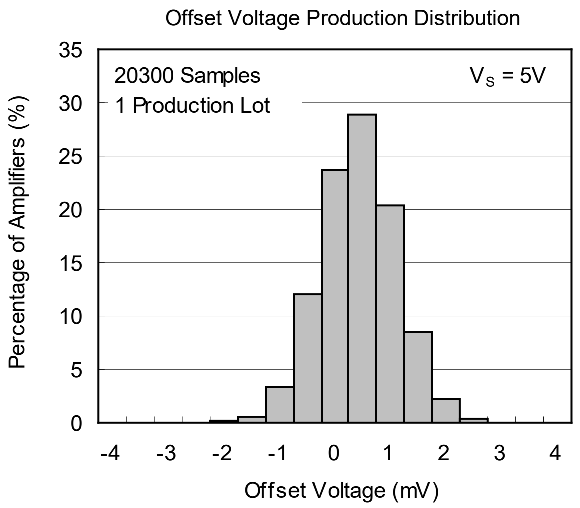


Output Voltage Swing vs. Output Current



TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At T_A = +25℃, V_{CM} = V_S/2, R_L = 600Ω, unless otherwise noted.



APPLICATION NOTES

Driving Capacitive Loads

The SGM8632C can directly drive 1000pF in unity-gain without oscillation. The unity-gain follower (buffer) is the most sensitive configuration to capacitive loading. Direct capacitive loading reduces the phase margin of the amplifier and this results in ringing or even oscillation. Applications that require greater capacitive driving capability should use an isolation resistor between the output and the capacitive load like the circuit in Figure 1. The isolation resistor R_{ISO} and the load capacitor C_L form a zero to increase stability. The bigger the R_{ISO} resistor value, the more stable V_{OUT} will be. Note that this method results in a loss of gain accuracy because R_{ISO} forms a voltage divider with the R_{LOAD}.

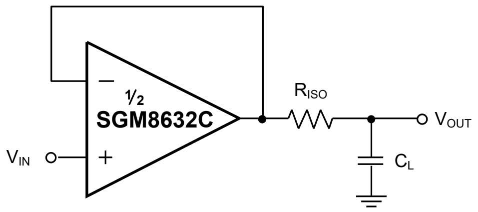


Figure 1. Indirectly Driving Heavy Capacitive Load

An improved circuit is shown in Figure 2. It provides DC accuracy as well as AC stability. R_F provides the DC accuracy by connecting the inverting signal with the output. C_F and R_{ISO} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

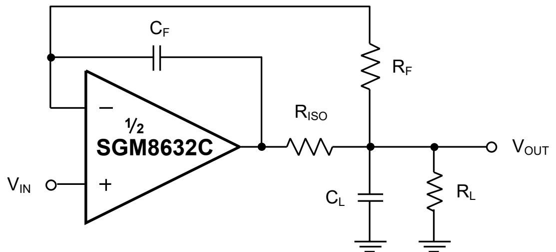


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For non-buffer configuration, there are two other ways to increase the phase margin: (a) by increasing the amplifier's closed-loop gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

Power-Supply Bypassing and Layout

The SGM8632C operates from either a single +2V to +5.5V supply or dual ±1V to ±2.75V supplies. For single-supply operation, bypass the power supply $+V_S$ with a 0.1µF ceramic capacitor which should be placed close to the $+V_S$ pin. For dual-supply operation, both the $+V_S$ and the $-V_S$ supplies should be bypassed to ground with separate 0.1µF ceramic capacitors. 2.2µF tantalum capacitor can be added for better performance.

Good PC board layout techniques optimize performance by decreasing the amount of stray capacitance at the op amp's inputs and output. To decrease stray capacitance, minimize trace lengths and widths by placing external components as close to the device as possible. Use surface-mount components whenever possible.

For the operational amplifier, soldering the part to the board directly is strongly recommended. Try to keep the high frequency current loop area small to minimize the EMI (electromagnetic interference).

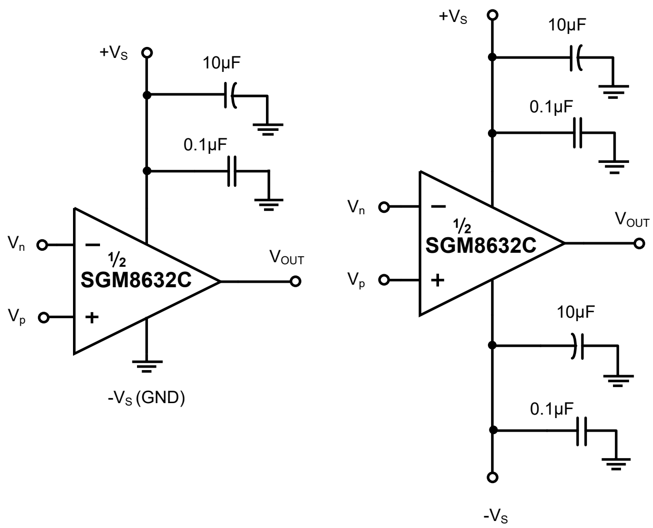


Figure 3. Amplifier with Bypass Capacitors

Grounding

A ground plane layer is important for SGM8632C circuit design. The length of the current path in an inductive ground return will create an unwanted voltage noise. Broad ground plane areas will reduce the parasitic inductance.

Input-to-Output Coupling

To minimize capacitive coupling, the input and output signal traces should not be in parallel. This helps reduce unwanted positive feedback.

TYPICAL APPLICATION CIRCUITS

Differential Amplifier

The circuit shown in Figure 4 performs the difference function. If the resistor ratios are equal ($R_4/R_3 = R_2/R_1$), then $V_{OUT} = (V_p - V_n) \times R_2/R_1 + V_{REF}$.

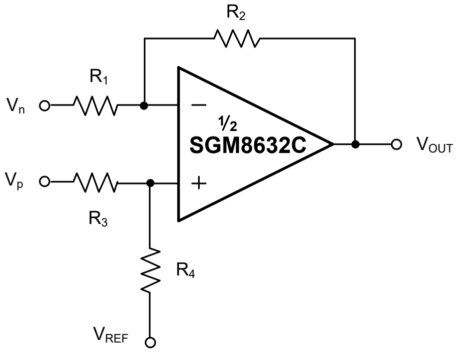


Figure 4. Differential Amplifier

Instrumentation Amplifier

The circuit in Figure 5 performs the same function as that in Figure 4 but with a high input impedance.

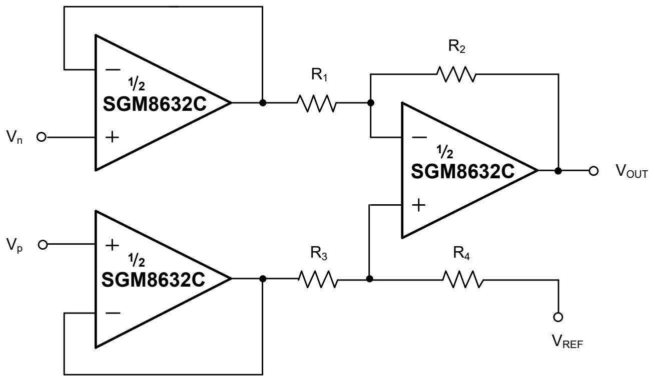


Figure 5. Instrumentation Amplifier

Active Low-Pass Filter

The low-pass filter shown in Figure 6 has a DC gain of ($-R_2/R_1$) and the -3dB corner frequency is $1/2\pi R_2C$. Make sure the filter bandwidth is within the bandwidth of the amplifier. Feedback resistors with large values can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistor values as low as possible and consistent with output loading consideration.

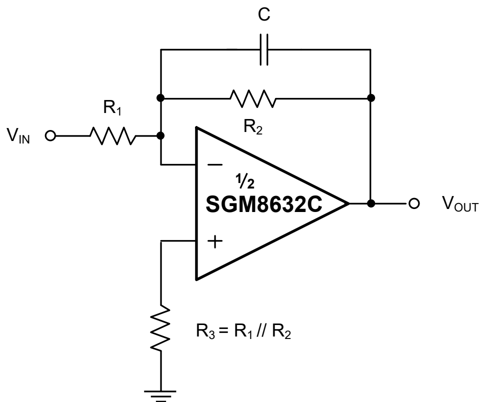


Figure 6. Active Low-Pass Filter

REVISION HISTORY

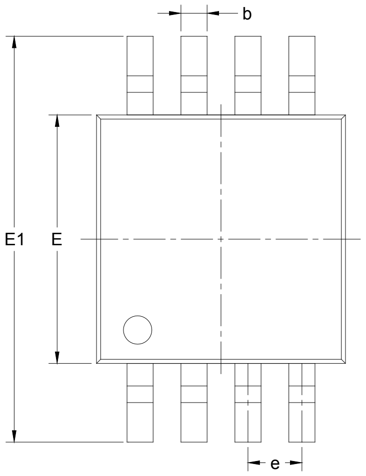
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

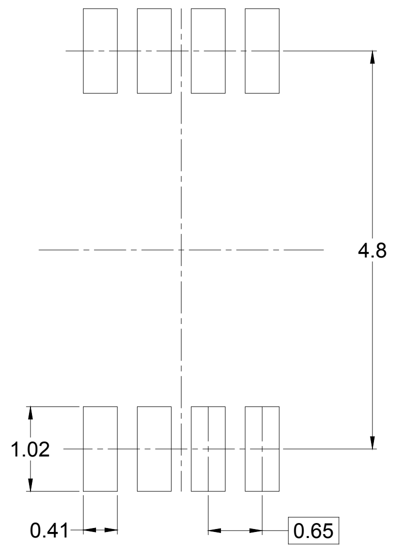
Changes from Original (NOVEMBER 2017) to REV.A

Changed from product preview to production data .... All

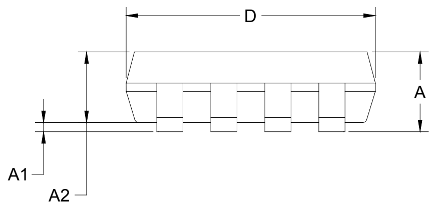
PACKAGE OUTLINE DIMENSIONS

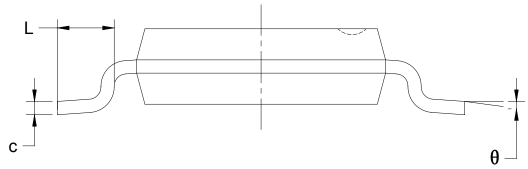
MSOP-8





RECOMMENDED LAND PATTERN (Unit: mm)

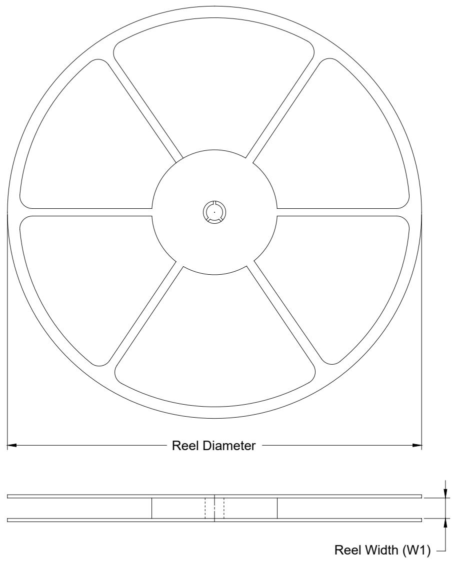




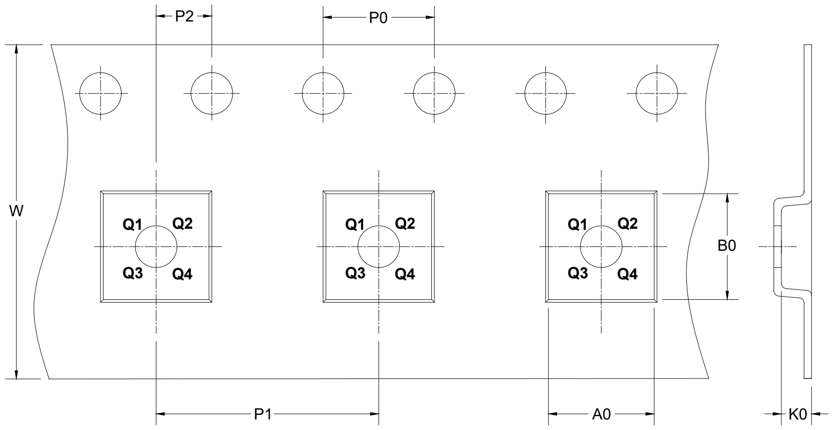
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.006
A2	0.750	0.950	0.030	0.037
b	0.250	0.380	0.010	0.015
c	0.090	0.230	0.004	0.009
D	2.900	3.100	0.114	0.122
E	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
e	0.650 BSC		0.026 BSC	
L	0.400	0.800	0.016	0.031
θ	0°	6°	0°	6°

TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



DIRECTION OF FEED

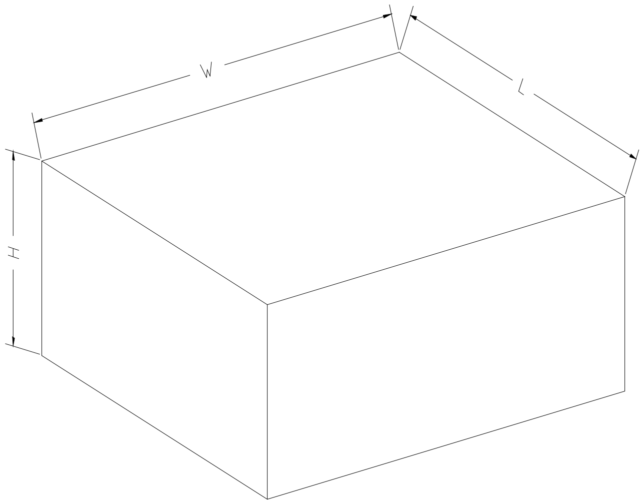
NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
MSOP-8	13″	12.4	5.20	3.30	1.50	4.0	8.0	2.0	12.0	Q1

DD0001

CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF CARTON BOX

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton
13″	386	280	370	5

DD0002