OPA379
OPA2379
OPA4379

1.8V, 2.9μA, 90kHz, Rail-to-Rail I/O OPERATIONAL AMPLIFIERS

FEATURES

- **LOW NOISE: 2.8μV_{PP} (0.1Hz - 10Hz)**
- ***micro*Power: 5.5μA (max)**
- **LOW OFFSET VOLTAGE: 1.5mV (max)**
- **DC PRECISION:**
  - **CMRR: 100dB**
  - **PSRR: 2μV/V**
  - **A_{OL}: 120dB**
- **WIDE SUPPLY VOLTAGE RANGE: 1.8V to 5.5V**
- ***micro*Size PACKAGES:**
  - **SC70-5, SOT23-5, SOT23-8, SO-8, TSSOP-14**

APPLICATIONS

- **BATTERY-POWERED INSTRUMENTS**
- **PORTABLE DEVICES**
- **MEDICAL INSTRUMENTS**
- **HANDHELD TEST EQUIPMENT**

DESCRIPTION

The OPA379 family of micropower, low-voltage operational amplifiers is designed for battery-powered applications. These amplifiers operate on a supply voltage as low as 1.8V (±0.9V). High-performance, single-supply operation with rail-to-rail capability (10μV max) makes the OPA379 family useful for a wide range of applications.

In addition to *micro*Size packages, the OPA379 family of op amps features impressive bandwidth (90kHz), low bias current (5pA), and low noise (80nV/√Hz) relative to the very low quiescent current (5.5μA max).

The OPA379 (single) is available in SC70-5, SOT23-5, and SO-8 packages. The OPA2379 (dual) comes in SOT23-8 and SO-8 packages. The OPA4379 (quad) is offered in a TSSOP-14 package. All versions are specified from –40°C to +125°C.

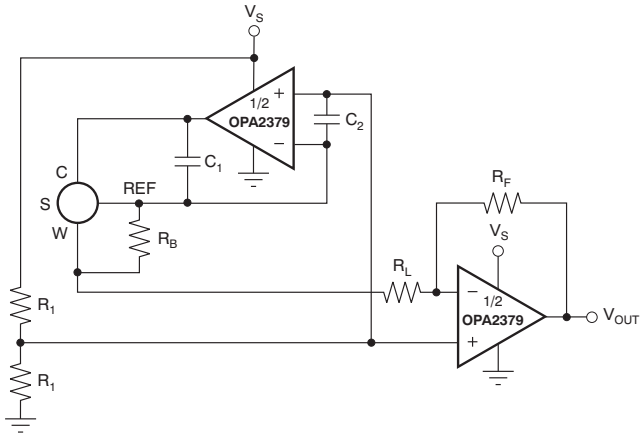


Figure 1. OPA2379 in Portable Gas Meter Application

Table 1. OPAx379 RELATED PRODUCTS

FEATURES	PRODUCT
1μA, 70kHz, 2mV V_{OS}, 1.8V to 5.5V Supply	OPAx349
1μA, 5.5kHz, 390μV V_{OS}, 2.5V to 16V Supply	TLV240x
1μA, 5.5kHz, 0.6mV V_{OS}, 2.5V to 12V Supply	TLV224x
7μA, 160kHz, 0.5mV V_{OS}, 2.7V to 16V Supply	TLV27Lx
7μA, 160kHz, 0.5mV V_{OS}, 2.7V to 16V Supply	TLV238x
20μA, 350kHz, 2mV V_{OS}, 2.3V to 5.5V Supply	OPAx347
20μA, 500kHz, 550μV V_{OS}, 1.8V to 3.6V Supply	TLV276x
45μA, 1MHz, 1mV V_{OS}, 2.1V to 5.5V Supply	OPAx348

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ABSOLUTE MAXIMUM RATINGS(1)

Over operating free-air temperature range (unless otherwise noted).

			OPA379, OPA2379, OPA4379	UNIT
Supply Voltage		V_S = (V+) – (V–)	+7	V
Signal Input Terminals, Voltage(2)			(V–) – 0.5 to (V+) + 0.5	V
Signal Input Terminals, Current(2)			±10	mA
Output Short-Circuit(3)			Continuous	
Operating Temperature		T_A	–40 to +125	°C
Storage Temperature		T_A	–65 to +150	°C
Junction Temperature		T_J	+150	°C
ESD Rating	Human Body Model	(HBM)	2000	V
	Charged Device Model	(CDM)	1000	V

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not supported.
(2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.
(3) Short-circuit to ground, one amplifier per package.

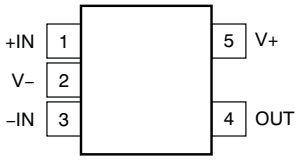
PACKAGE/ORDERING INFORMATION(1)

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	PACKAGE MARKING
OPA379	SC70−5	DCK	AYR
	SOT23−5	DBV	B53
	SO−8	D	OPA379A
OPA2379	SOT23−8	DCN	B61
	SO−8	D	OPA2379A
OPA4379	TSSOP−14	PW	OPA4379A

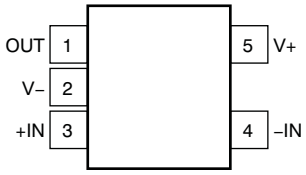
(1) For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

PIN CONFIGURATIONS

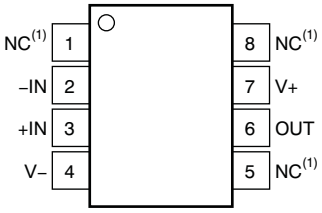
OPA379
SC70-5
(TOP VIEW)



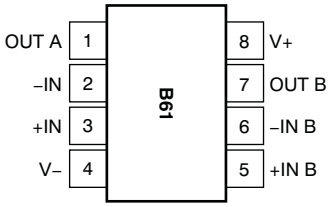
OPA379
SOT23-5
(TOP VIEW)



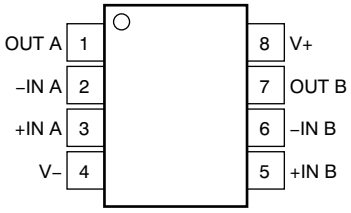
OPA379
SO-8
(TOP VIEW)



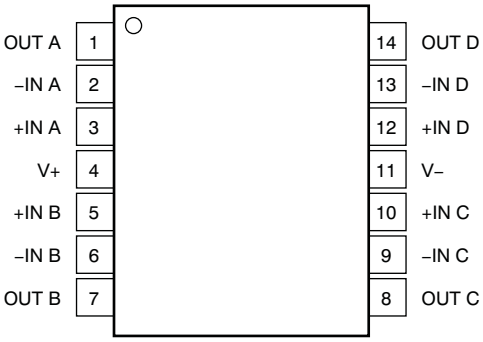
OPA2379[(2)]
SOT23-8
(TOP VIEW)



OPA2379
SO-8
(TOP VIEW)



OPA4379
TSSOP-14
(TOP VIEW)



(1) NC denotes no internal connection.

(2) Pin 1 of the SOT23−8 package is determined by orienting the package marking as shown.

ELECTRICAL CHARACTERISTICS: V_S = +1.8V to +5.5V

Boldface limits apply over the specified temperature range indicated.
At T_A = +25°C, R_L = 25kΩ connected to V_S/2, and V_{CM} < (V+) – 1V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	OPA379, OPA2379, OPA4379 MIN	TYP	MAX	UNIT
OFFSET VOLTAGE						
Initial Offset Voltage	V_{OS}	V_S = 5V		0.4	1.5	mV
Over –40°C to +125°C					**2**	**mV**
Drift, –40°C to +85°C	**dV_{OS}/dT**			**1.5**		**µV/°C**
Drift, –40°C to +125°C				**2.7**		**µV/°C**
vs Power Supply	PSRR			2	10	µV/V
Over –40°C to +125°C					**20**	**µV/V**
INPUT VOLTAGE RANGE						
Common-Mode Voltage Range	V_{CM}		(V–) – 0.1 to (V+) + 0.1			V
Common-Mode Rejection Ratio(1)	CMRR	(V–) < V_{CM} < (V+) – 1V	90	100		dB
Over –40°C to +85°C		**(V–) < V_{CM} < (V+) – 1V**	**80**			**dB**
Over –40°C to +125°C		**(V–) < V_{CM} < (V+) – 1V**	**62**			**dB**
INPUT BIAS CURRENT						
Input Bias Current	I_B	V_S = 5V, V_{CM} ≤ V_S/2		±5	±50	pA
Input Offset Current	I_{OS}	V_S = 5V		±5	±50	pA
INPUT IMPEDANCE						
Differential				10^{13} || 3		Ω || pF
Common-Mode				10^{13} || 6		Ω || pF
NOISE						
Input Voltage Noise		f = 0.1Hz to 10Hz		2.8		$µV_{PP}$
Input Voltage Noise Density	e_n	f = 1kHz		80		nV/√Hz
Input Current Noise Density	i_n	f = 1kHz		1		fA/√Hz
OPEN-LOOP GAIN						
Open-Loop Voltage Gain	A_{OL}	V_S = 5V, R_L = 25kΩ, 100mV < V_O < (V+) – 100mV	100	120		dB
Over –40°C to +125°C		**V_S = 5V, R_L = 25kΩ, 100mV < V_O < (V+) – 100mV**	**80**			**dB**
		V_S = 5V, R_L = 5kΩ, 500mV < V_O < (V+) – 500mV	100	120		dB
Over –40°C to +125°C		**V_S = 5V, R_L = 5kΩ, 500mV < V_O < (V+) – 500mV**	**80**			**dB**
OUTPUT						
Voltage Output Swing from Rail		R_L = 25kΩ		5	10	mV
Over –40°C to +125°C		**R_L = 25kΩ**			**15**	**mV**
		R_L = 5kΩ		25	50	mV
Over –40°C to +125°C		**R_L = 5kΩ**			**75**	**mV**
Short-Circuit Current	I_{SC}			±5		mA
Capacitive Load Drive	C_{LOAD}		See Typical Characteristics			
Closed-Loop Output Impedance	R_{OUT}	G = 1, f = 1kHz, I_O = 0		10		Ω
Open-Loop Output Impedance	R_O	f = 100kHz, I_O = 0		28		kΩ
FREQUENCY RESPONSE		C_{LOAD} = 30pF				
Gain Bandwidth Product	GBW			90		kHz
Slew Rate	SR	G = +1		0.03		V/µs
Overload Recovery Time		V_{IN} × GAIN > V_S		25		µs
Turn-On Time	t_{ON}			1		ms

(1) See Typical Characteristic gragh, *Common-Mode Rejection Ratio vs Frequency* (Figure 3).

ELECTRICAL CHARACTERISTICS: V_S = +1.8V to +5.5V (continued)

Boldface limits apply over the specified temperature range indicated.
At T_A = +25°C, R_L = 25kΩ connected to V_S/2, and V_{CM} < (V+) – 1V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	OPA379, OPA2379, OPA4379			UNIT
			MIN	TYP	MAX	
POWER SUPPLY						
Specified/Operating Voltage Range	V_S		1.8		5.5	V
Quiescent Current per Amplifier	I_Q	V_S = 5.5V, I_O = 0		2.9	5.5	µA
Over –40°C to +125°C					**10**	µ**A**
TEMPERATURE						
Specified/Operating Range	T_A		–40		+125	°C
Storage Range	T_J		–65		+150	°C
Thermal Resistance	θ_{JA}					
SC70−5				250		°C/W
SOT23−5				200		°C/W
SOT23−8, TSSOP−14, SO−8				150		°C/W

TYPICAL CHARACTERISTICS

At T_A = +25°C, V_S = 5V, and R_L = 25kΩ connected to V_S/2, unless otherwise noted.

OPEN-LOOP GAIN AND PHASE vs FREQUENCY

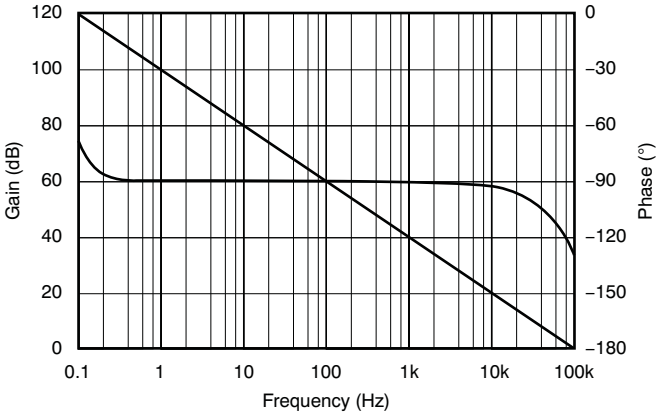


Figure 2.

COMMON-MODE AND POWER-SUPPLY REJECTION RATIO vs FREQUENCY

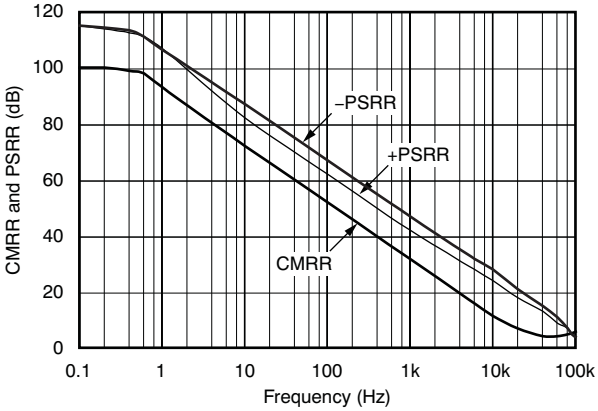


Figure 3.

MAXIMUM OUTPUT VOLTAGE vs FREQUENCY

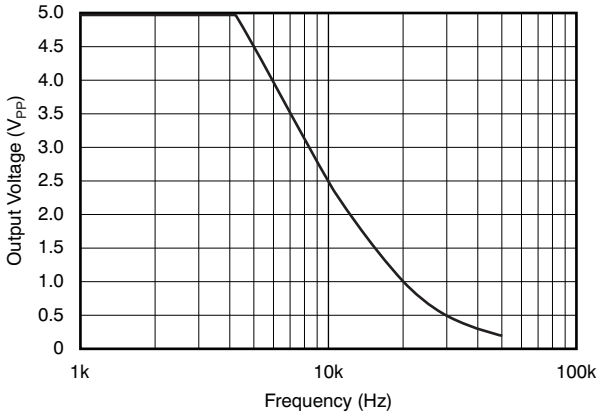


Figure 4.

QUIESCENT CURRENT vs SUPPLY VOLTAGE

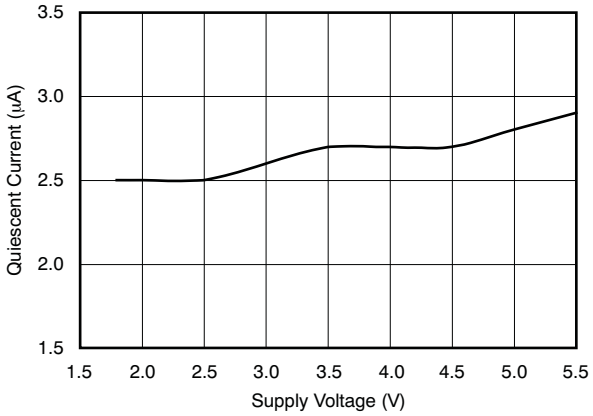


Figure 5.

OUTPUT VOLTAGE vs OUTPUT CURRENT

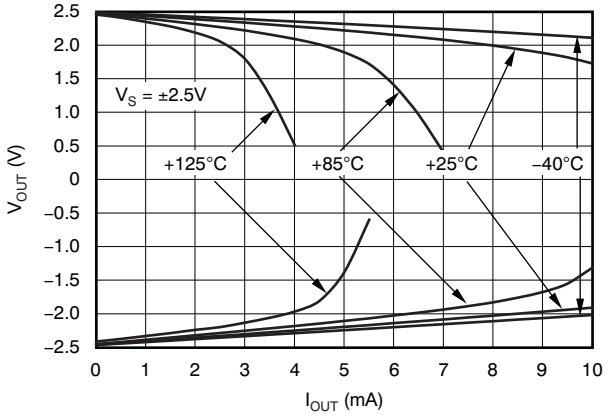


Figure 6.

SHORT-CIRCUIT CURRENT vs SUPPLY VOLTAGE

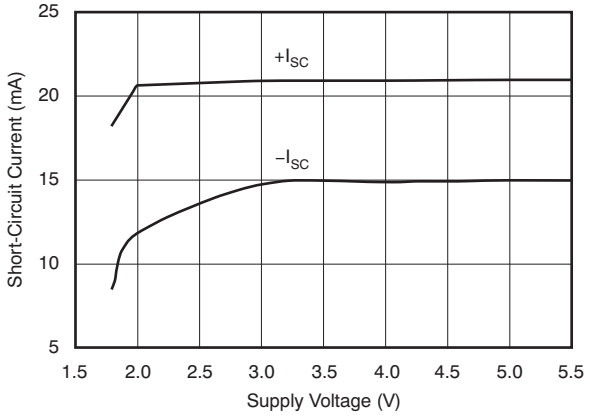


Figure 7.

TYPICAL CHARACTERISTICS (continued)

At T_A = +25°C, V_S = 5V, and R_L = 25kΩ connected to V_S/2, unless otherwise noted.

OFFSET VOLTAGE vs COMMON-MODE VOLTAGE vs TEMPERATURE

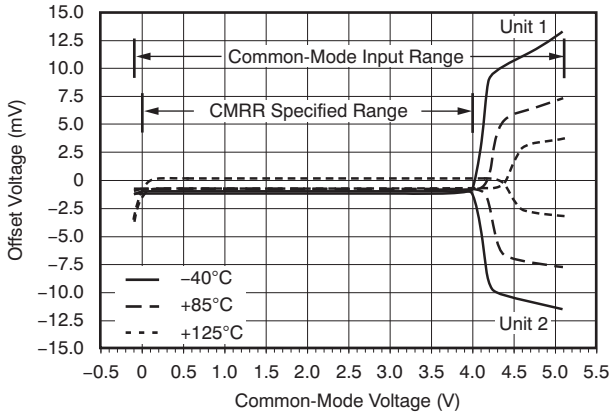


Figure 8.

OFFSET VOLTAGE PRODUCTION DISTRIBUTION

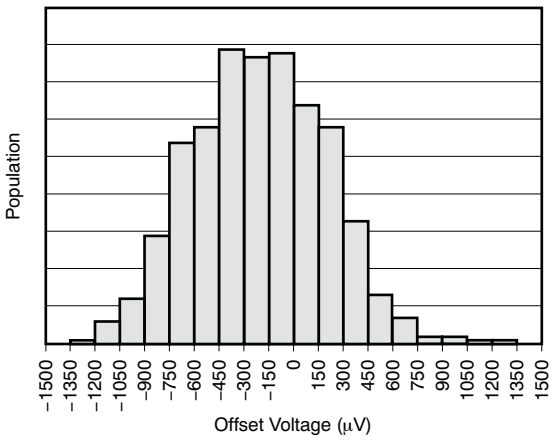


Figure 9.

OFFSET VOLTAGE DRIFT DISTRIBUTION (−40°C to +85°C)

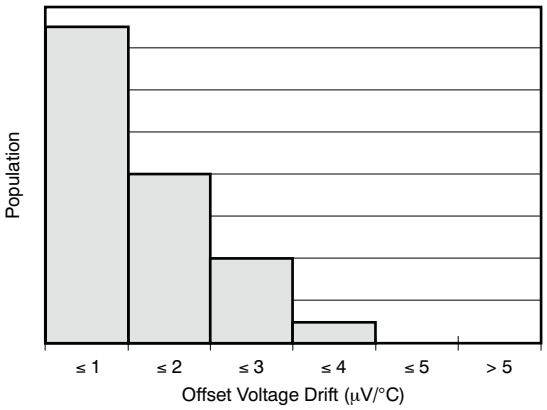


Figure 10.

OFFSET VOLTAGE DRIFT DISTRIBUTION (−40°C to +125°C)

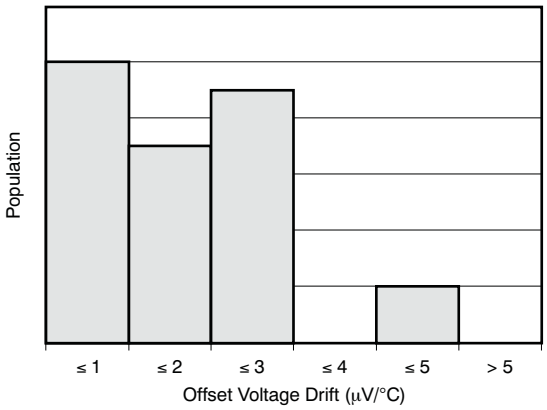


Figure 11.

QUIESCENT CURRENT vs TEMPERATURE

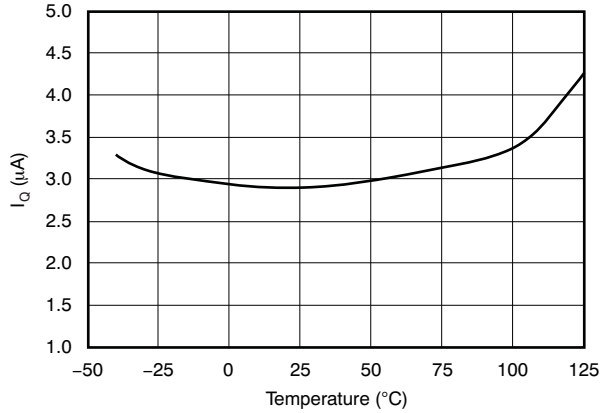


Figure 12.

QUIESCENT CURRENT PRODUCTION DISTRIBUTION

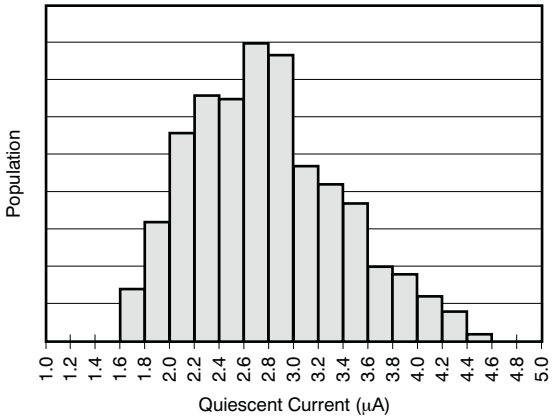


Figure 13.

TYPICAL CHARACTERISTICS (continued)

At T_A = +25°C, V_S = 5V, and R_L = 25kΩ connected to V_S/2, unless otherwise noted.

INPUT BIAS CURRENT vs TEMPERATURE

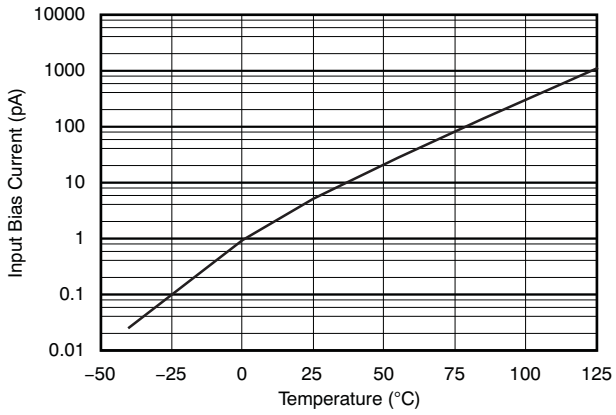


Figure 14.

0.1Hz TO 10Hz NOISE

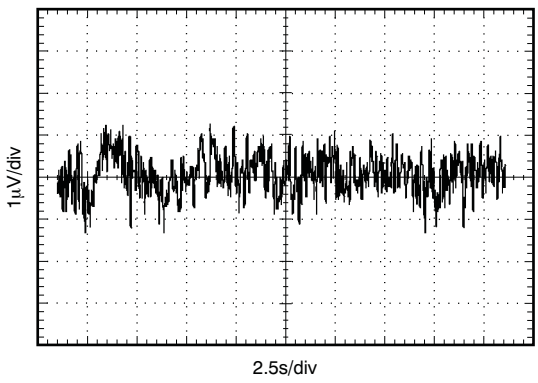


Figure 15.

NOISE vs FREQUENCY

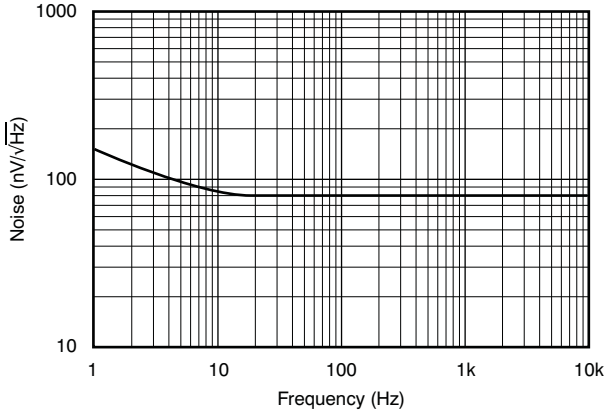


Figure 16.

SMALL-SIGNAL OVERSHOOT vs CAPACITIVE LOAD

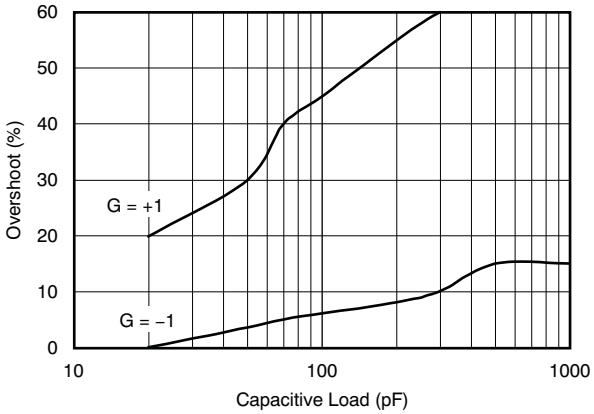


Figure 17.

SMALL-SIGNAL STEP RESPONSE

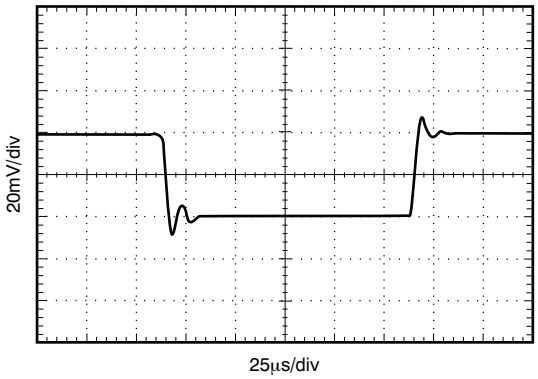


Figure 18.

LARGE-SIGNAL STEP RESPONSE

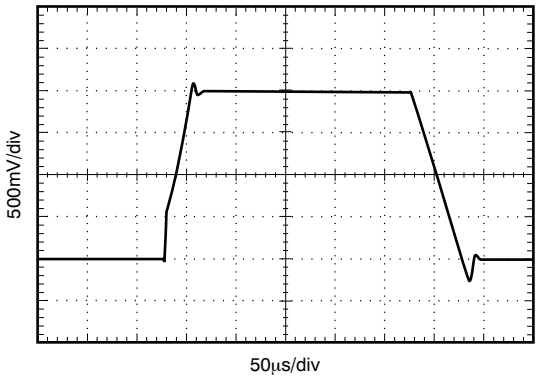


Figure 19.

APPLICATION INFORMATION

The OPA379 family of operational amplifiers minimizes power consumption without compromising bandwidth or noise. Power-supply rejection ratio (PSRR), common-mode rejection ratio (CMRR), and open-loop gain (A_{OL}) typical values are 100dB or better.

When designing for ultra-low power, choose system components carefully. To minimize current consumption, select large-value resistors. Any resistors will react with stray capacitance in the circuit and the input capacitance of the operational amplifier. These parasitic RC combinations can affect the stability of the overall system. A feedback capacitor may be required to assure stability and limit overshoot or gain peaking.

Good layout practice mandates the use of a 0.1µF bypass capacitor placed closely across the supply pins.

OPERATING VOLTAGE

OPA379 series op amps are fully specified and tested from +1.8V to +5.5V (±0.9V to ±2.75V). Parameters that will vary with supply voltage are shown in the Typical Characteristics curves.

INPUT COMMON-MODE VOLTAGE RANGE

The input common-mode voltage range of the OPA379 family typically extends 100mV beyond each supply rail. This rail-to-rail input is achieved using a complementary input stage. CMRR is specified from the negative rail to 1V below the positive rail. Between (V+) – 1V and (V+) + 0.1V, the amplifier operates with higher offset voltage because of the transition region of the input stage. See the typical characteristic, *Offset Voltage vs Common-Mode Voltage vs Temperature* (Figure 8).

PROTECTING INPUTS FROM OVER-VOLTAGE

Normally, input currents are 5pA. However, a large voltage input (greater than 500mV beyond the supply rails) can cause excessive current to flow in or out of the input pins. Therefore, as well as keeping the input voltage below the maximum rating, it is also important to limit the input current to less than 10mA. This limiting is easily accomplished with an input voltage resistor, as shown in Figure 20.

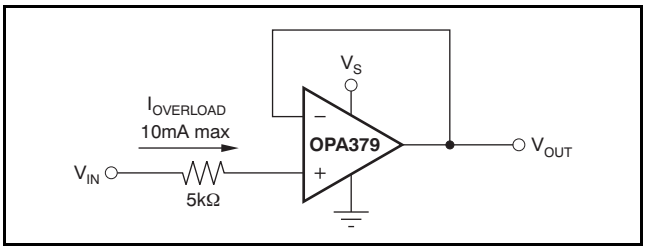


Figure 20. Input Current Protection for Voltages Exceeding the Supply Voltage

NOISE

Although micropower amplifiers frequently have high wideband noise, the OPA379 series offer excellent noise performance. Resistors should be chosen carefully because the OPA379 has only 2.8µV_{PP} of 0.1Hz to 10Hz noise, and 80nV/$\sqrt{Hz}$ of wideband noise; otherwise, they can become the dominant source of noise.

CAPACITIVE LOAD AND STABILITY

Follower configurations with load capacitance in excess of 30pF can produce extra overshoot (see typical characteristic *Small-Signal Overshoot vs Capacitive Load*, Figure 17) and ringing in the output signal. Increasing the gain enhances the ability of the amplifier to drive greater capacitive loads. In unity-gain configurations, capacitive load drive can be improved by inserting a small (10Ω to 20Ω) resistor, R_S, in series with the output, as shown in Figure 21. This resistor significantly reduces ringing while maintaining direct current (dc) performance for purely capacitive loads. However, if there is a resistive load in parallel with the capacitive load, a voltage divider is created, introducing a dc error at the output and slightly reducing the output swing. The error introduced is proportional to the ratio R_S/R_L, and is generally negligible.

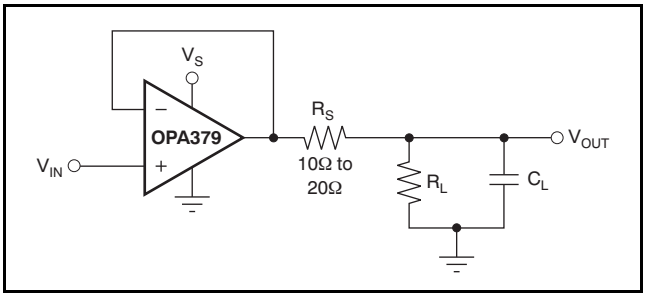


Figure 21. Series Resistor in Unity-Gain Buffer Configuration Improves Capacitive Load Drive

In unity-gain inverter configuration, phase margin can be reduced by the reaction between the capacitance at the op amp input and the gain setting resistors. Best performance is achieved by using smaller valued resistors. However, when large valued resistors cannot be avoided, a small (4pF to 6pF) capacitor, C_{FB}, can be inserted in the feedback, as shown in Figure 22. This configuration significantly reduces overshoot by compensating the effect of capacitance, C_{IN}, which includes the amplifier input capacitance (3pf) and printed circuit board (PC) parasitic capacitance.

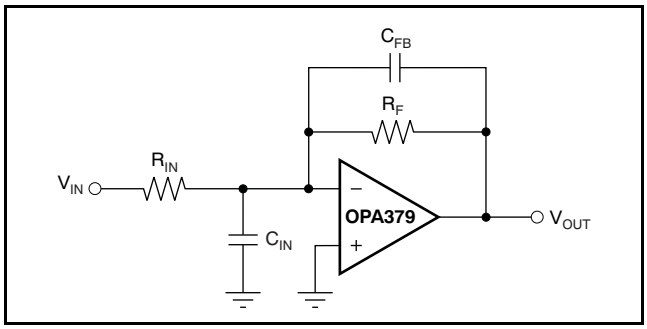


Figure 22. Improving Stability for Large R_F and R_{IN}

BATTERY MONITORING

The low operating voltage and quiescent current of the OPA379 series make it an excellent choice for battery monitoring applications, as shown in Figure 23. In this circuit, V_{STATUS} is high as long as the battery voltage remains above 2V. A low-power reference is used to set the trip point. Resistor values are selected as follows:

1. R_F Selecting: Select R_F such that the current through R_F is approximately 1000x larger than the maximum bias current over temperature:

$$R_F = \frac{V_{REF}}{1000(I_{BMAX})} = \frac{1.2V}{1000(100pA)} = 12M\Omega \approx 10M\Omega \quad (1)$$

2. Choose the hysteresis voltage, V_{HYST}. For battery monitoring applications, 50mV is adequate.
3. Calculate R_1 as follows:

$$R_1 = R_F \left[\frac{V_{HYST}}{V_{BATT}}\right] = 10M\Omega \left[\frac{50mW}{2.4V}\right] = 210k\Omega \quad (2)$$

4. Select a threshold voltage for V_{IN} rising (V_{THRS}) = 2.0V
5. Calculate R_2 as follows:

$$R_2 = \frac{1}{\left(\left[\frac{V_{THRS}}{V_{REF} \times R_1}\right] - \frac{1}{R_1} - \frac{1}{R_F}\right)} = \frac{1}{\left(\left[\frac{2V}{1.2V \times 210k\Omega}\right] - \frac{1}{210k\Omega} - \frac{1}{10M\Omega}\right)} = 325k\Omega \quad (3)$$

6. Calculate R_{BIAS}: The minimum supply voltage for this circuit is 1.8V. The REF1112 has a current requirement of 1.2µA (max). Providing 2µA of supply current assures proper operation. Therefore:

$$R_{BIAS} = \frac{(V_{BATTMIN} - V_{REF})}{I_{BIAS}} = \frac{(1.8V - 1.2V)}{2\mu A} = 0.3M\Omega \quad (4)$$

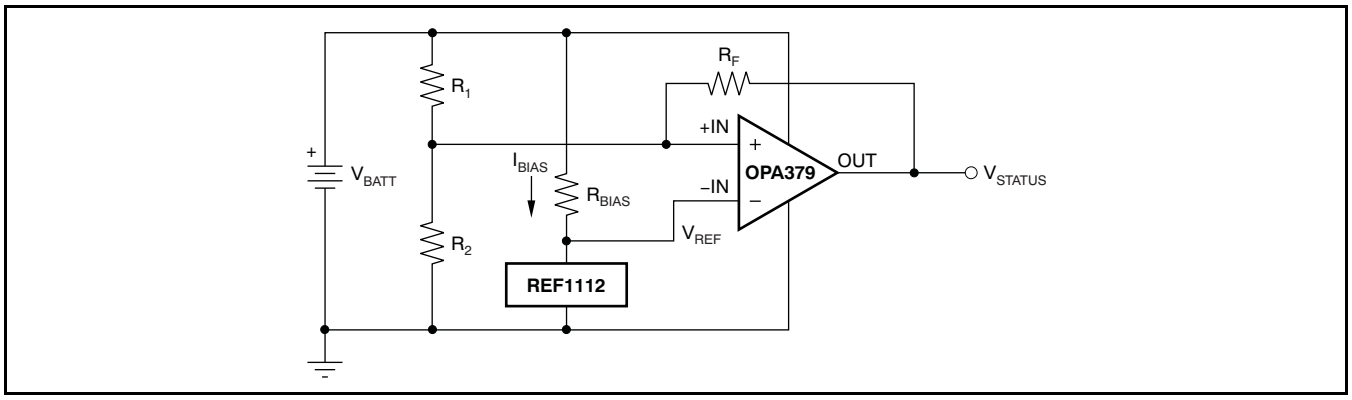


Figure 23. Battery Monitor

WINDOW COMPARATOR

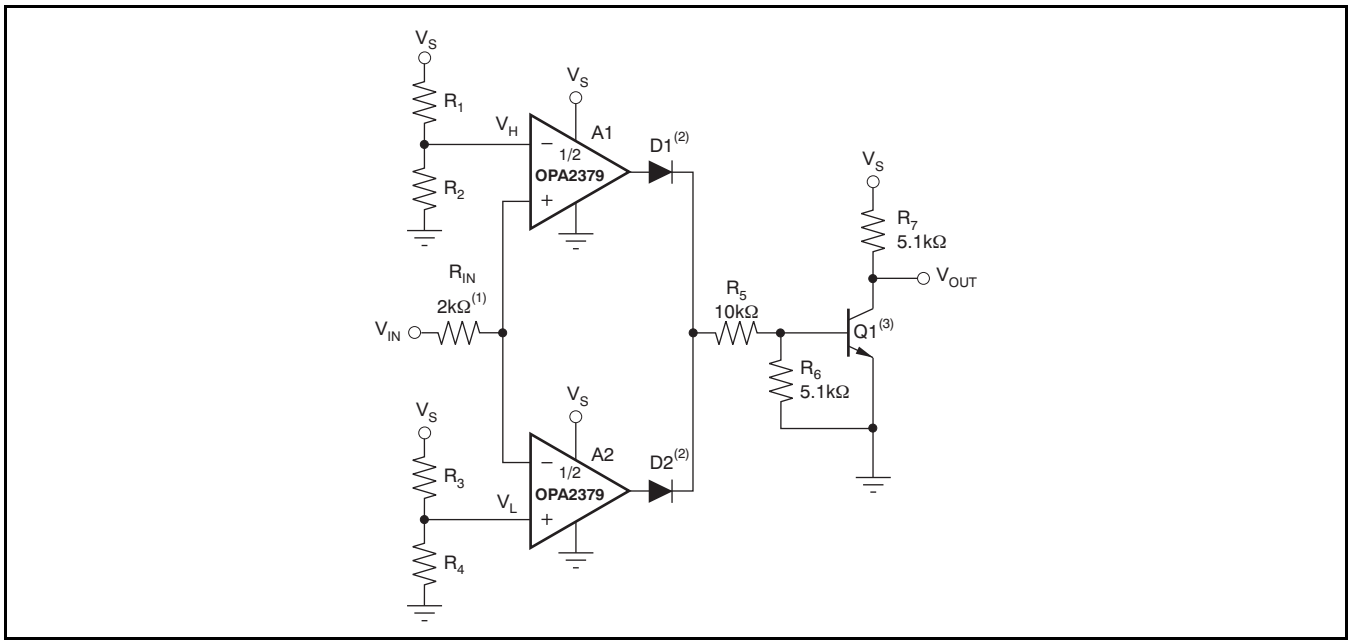
Figure 24 shows the OPA2379 used as a window comparator. The threshold limits are set by V_H and V_L, with $V_H > V_L$. When $V_{IN} < V_H$, the output of A1 is low. When $V_{IN} > V_L$, the output of A2 is low. Therefore, both op amp outputs are at 0V as long as V_{IN} is between V_H and V_L. This architecture results in no current flowing through either diode, Q1 in cutoff, with the base voltage at 0V, and V_{OUT} forced high.

If V_{IN} falls below V_L, the output of A2 is high, current flows through D2, and V_{OUT} is low. Likewise, if V_{IN} rises above V_H, the output of A1 is high, current flows through D1, and V_{OUT} is low.

The window comparator threshold voltages are set as follows:

$$V_H = \frac{R_2}{R_1 + R_2} \times V_S \quad (5)$$

$$V_L = \frac{R_4}{R_3 + R_4} \times V_S \quad (6)$$



(1) R_{IN} protects A1 and A2 from possible excess current flow.

(2) IN4446 or equivalent diodes.

(3) 2N2222 or equivalent NPN transistor.

Figure 24. OPA2379 as a Window Comparator

ADDITIONAL APPLICATION EXAMPLES

Figure 25 through Figure 29 illustrate additional application examples.

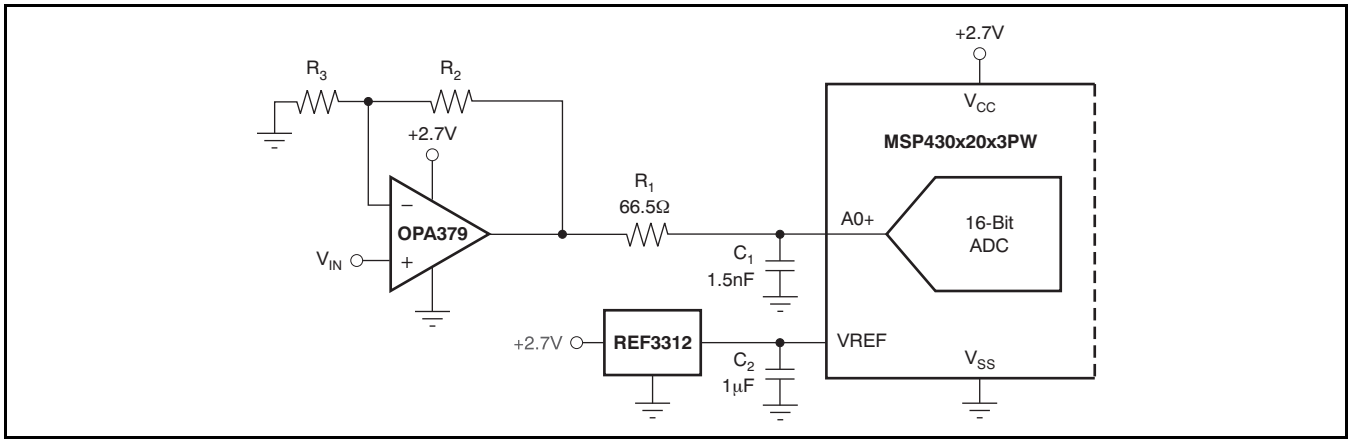


Figure 25. Unipolar Signal Chain Configuration

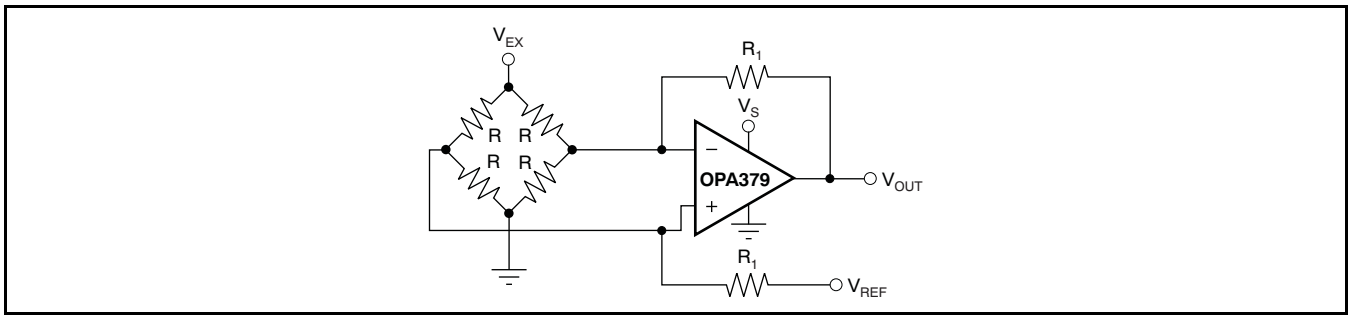
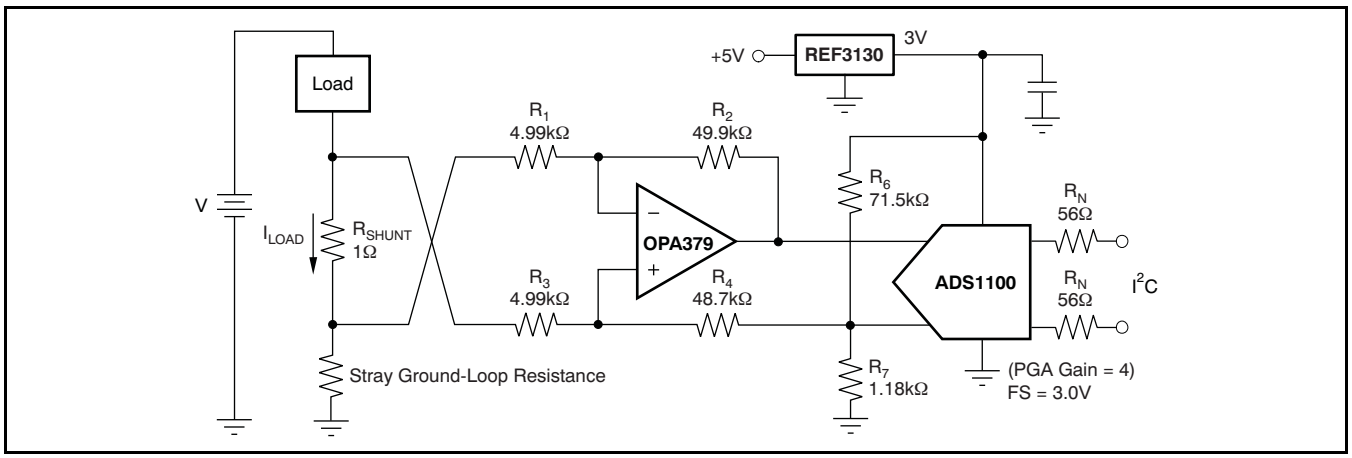
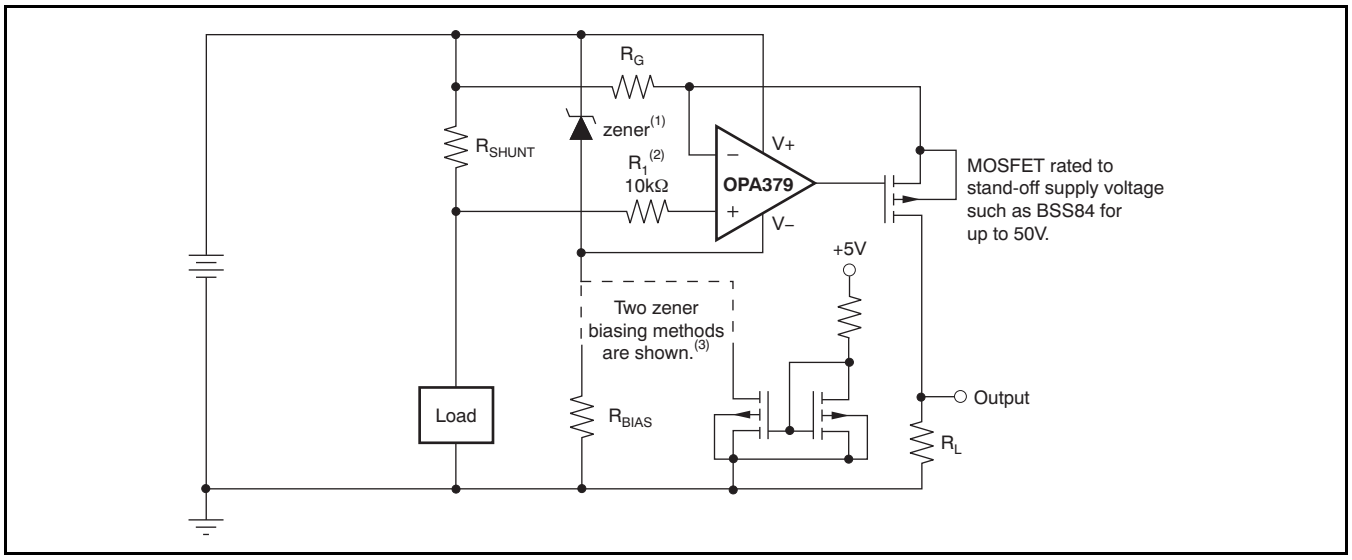


Figure 26. Single Op Amp Bridge Amplifier



NOTE: 1% resistors provide adequate common-mode rejection at small ground-loop errors.

Figure 27. Low-Side Current Monitor



(1) Zener rated for op amp supply capability (that is, 5.1V for OPA379).

(2) Current-limiting resistor.

(3) Choose zener biasing resistor or dual NMOSMETs (FDG6301N, NTJD4001N, or Si1034).

Figure 28. High-Side Current Monitor

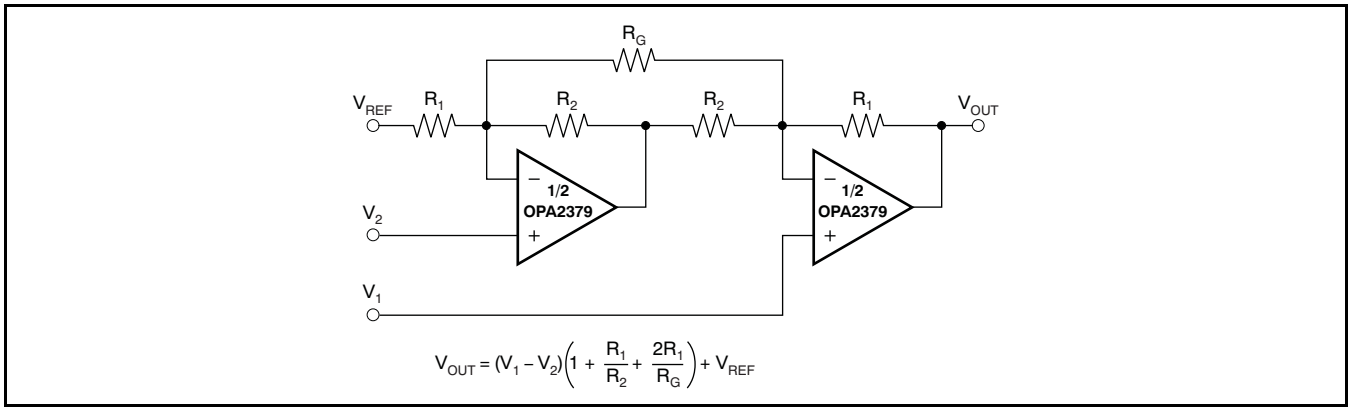


Figure 29. Two Op Amp Instrumentation Amplifier

PACKAGE OPTION ADDENDUM

15-Apr-2017

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA2379AID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2379A	Samples
OPA2379AIDCNR	ACTIVE	SOT-23	DCN	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	BPK	Samples
OPA2379AIDCNRG4	ACTIVE	SOT-23	DCN	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	BPK	Samples
OPA2379AIDCNT	ACTIVE	SOT-23	DCN	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	BPK	Samples
OPA2379AIDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2379A	Samples
OPA2379AIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2379A	Samples
OPA2379AIDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2379A	Samples
OPA379AID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 379A	Samples
OPA379AIDBVR	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B53	Samples
OPA379AIDBVRG4	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B53	Samples
OPA379AIDBVT	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B53	Samples
OPA379AIDBVTG4	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B53	Samples
OPA379AIDCKR	ACTIVE	SC70	DCK	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B54	Samples
OPA379AIDCKT	ACTIVE	SC70	DCK	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B54	Samples
OPA379AIDCKTG4	ACTIVE	SC70	DCK	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	B54	Samples
OPA379AIDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 379A	Samples
OPA379AIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 379A	Samples

PACKAGE OPTION ADDENDUM

15-Apr-2017

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA379AIDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 379A	Samples
OPA4379AIPWR	ACTIVE	TSSOP	PW	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	4379A	Samples
OPA4379AIPWRG4	ACTIVE	TSSOP	PW	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	4379A	Samples

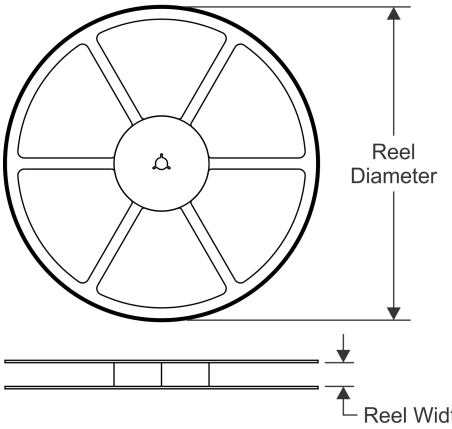
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

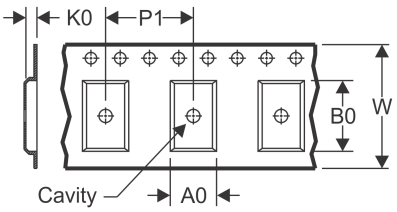
3-Aug-2017

TAPE AND REEL INFORMATION

REEL DIMENSIONS

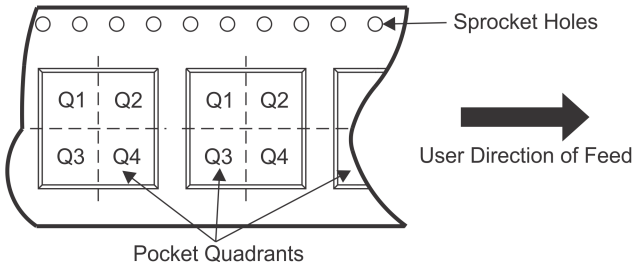


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



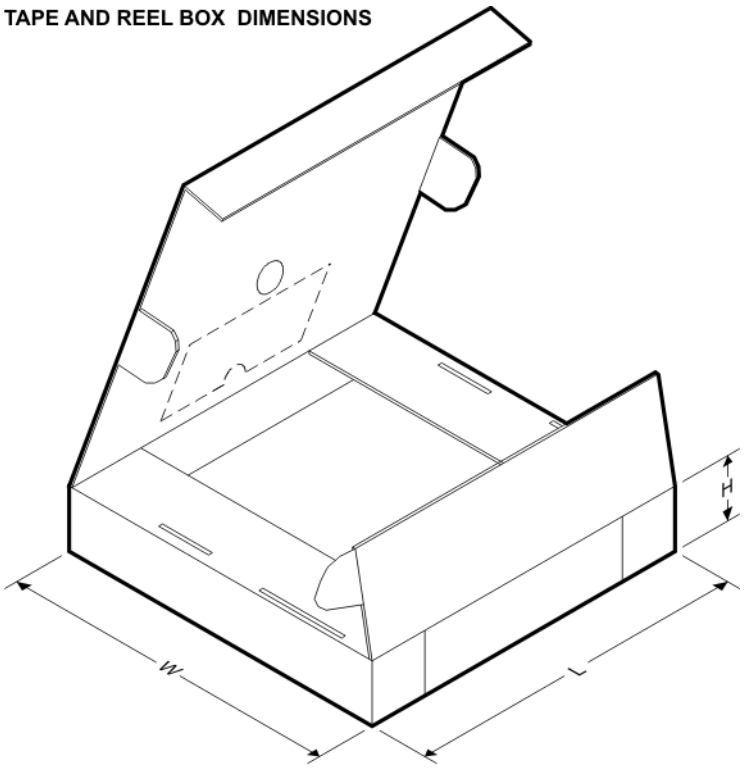
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2379AIDCNR	SOT-23	DCN	8	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA2379AIDCNT	SOT-23	DCN	8	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA2379AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA379AIDBVR	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA379AIDBVT	SOT-23	DBV	5	250	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA379AIDCKR	SC70	DCK	5	3000	179.0	8.4	2.2	2.5	1.2	4.0	8.0	Q3
OPA379AIDCKR	SC70	DCK	5	3000	178.0	9.0	2.4	2.5	1.2	4.0	8.0	Q3
OPA379AIDCKT	SC70	DCK	5	250	178.0	9.0	2.4	2.5	1.2	4.0	8.0	Q3
OPA379AIDCKT	SC70	DCK	5	250	179.0	8.4	2.2	2.5	1.2	4.0	8.0	Q3
OPA379AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA4379AIPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

PACKAGE MATERIALS INFORMATION

3-Aug-2017

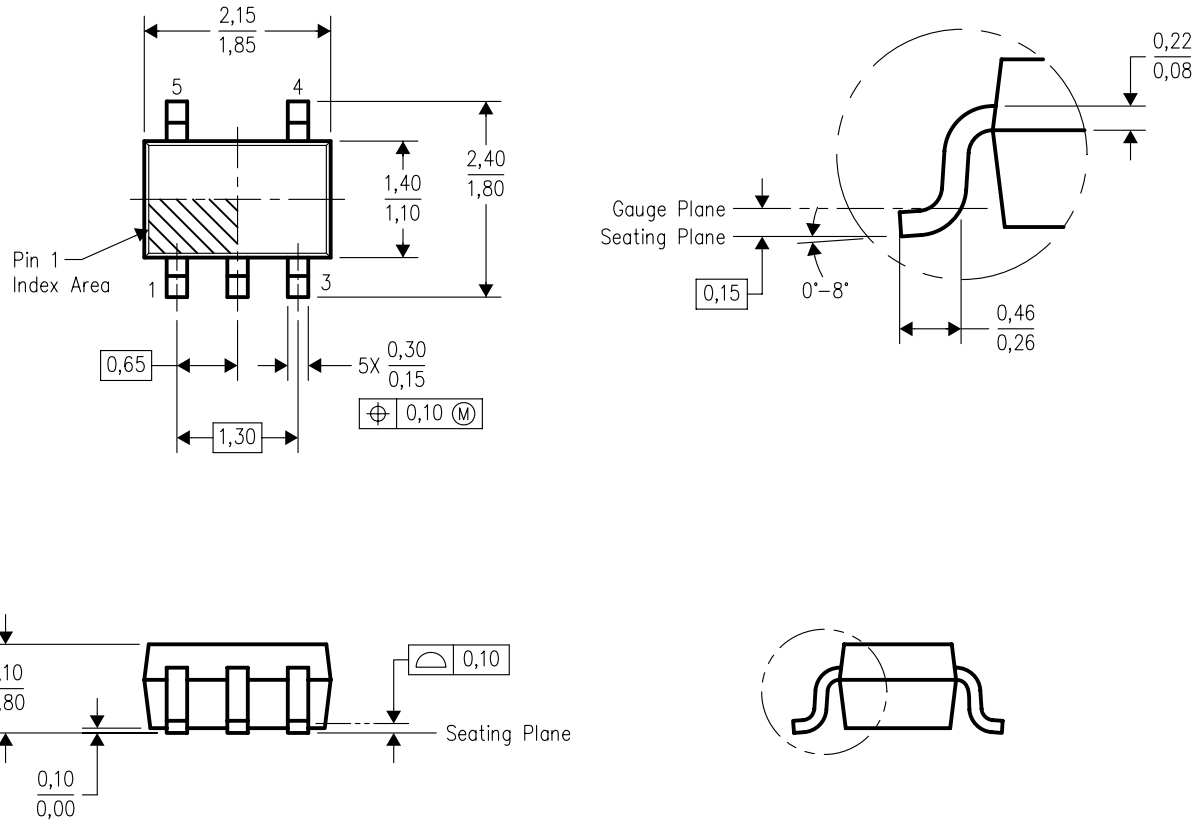
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2379AIDCNR	SOT-23	DCN	8	3000	195.0	200.0	45.0
OPA2379AIDCNT	SOT-23	DCN	8	250	195.0	200.0	45.0
OPA2379AIDR	SOIC	D	8	2500	367.0	367.0	35.0
OPA379AIDBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
OPA379AIDBVT	SOT-23	DBV	5	250	180.0	180.0	18.0
OPA379AIDCKR	SC70	DCK	5	3000	195.0	200.0	45.0
OPA379AIDCKR	SC70	DCK	5	3000	190.0	190.0	30.0
OPA379AIDCKT	SC70	DCK	5	250	190.0	190.0	30.0
OPA379AIDCKT	SC70	DCK	5	250	195.0	200.0	45.0
OPA379AIDR	SOIC	D	8	2500	367.0	367.0	35.0
OPA4379AIPWR	TSSOP	PW	14	2000	367.0	367.0	35.0

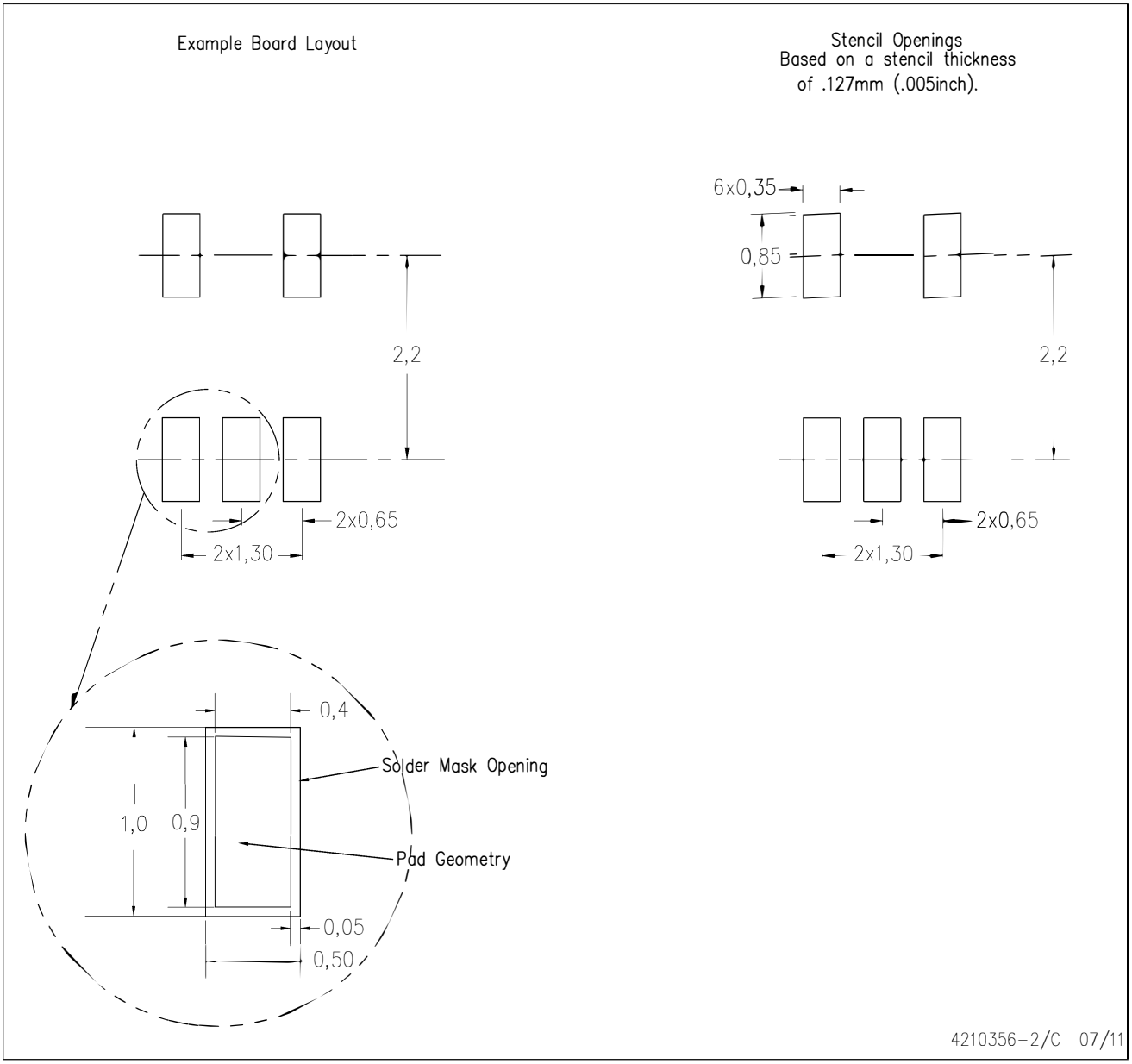
DCK (R–PDSO–G5) PLASTIC SMALL–OUTLINE PACKAGE



4093553–3/G 01/2007

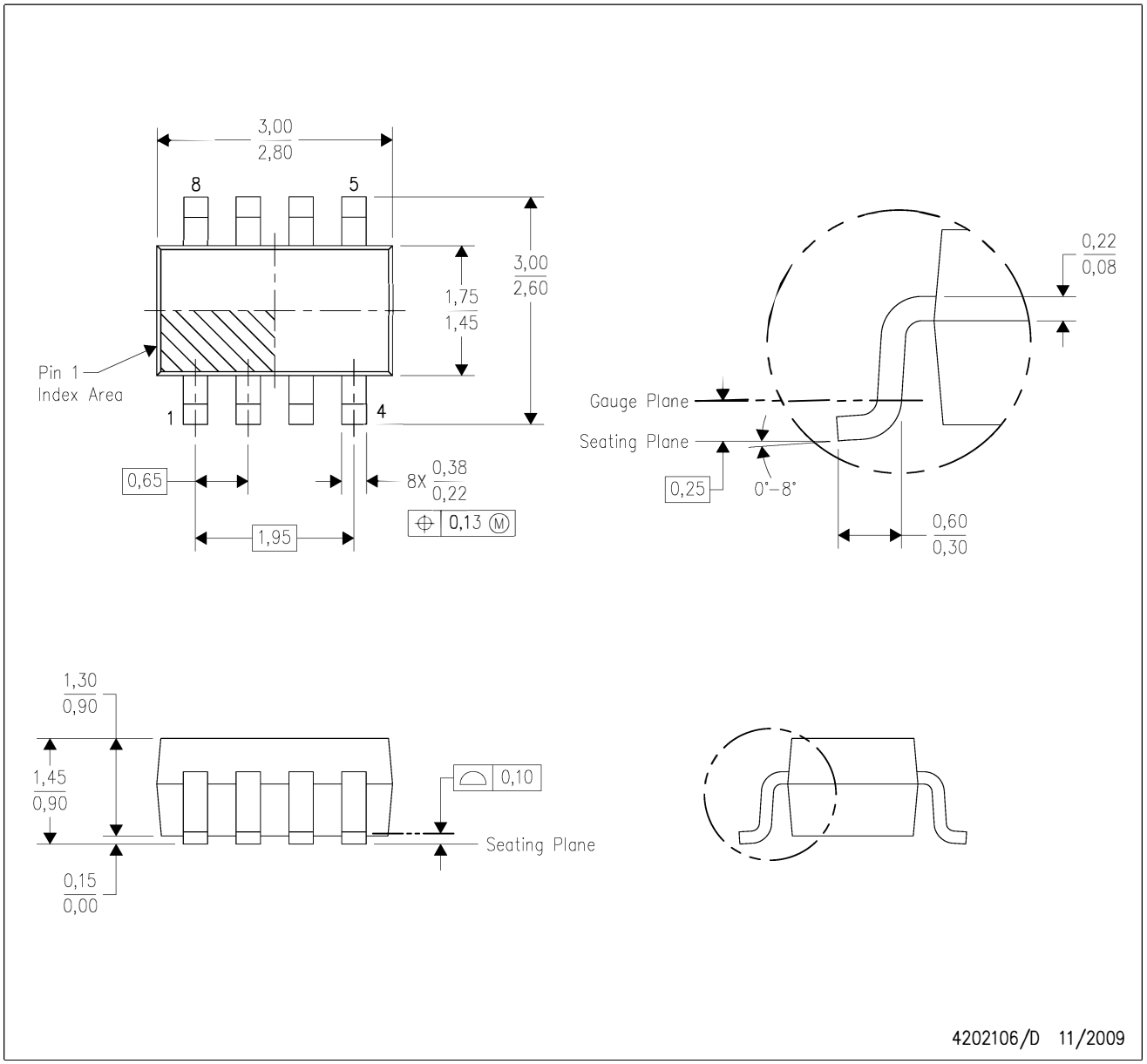
NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
D. Falls within JEDEC MO–203 variation AA.

DCK (R–PDSO–G5) PLASTIC SMALL OUTLINE



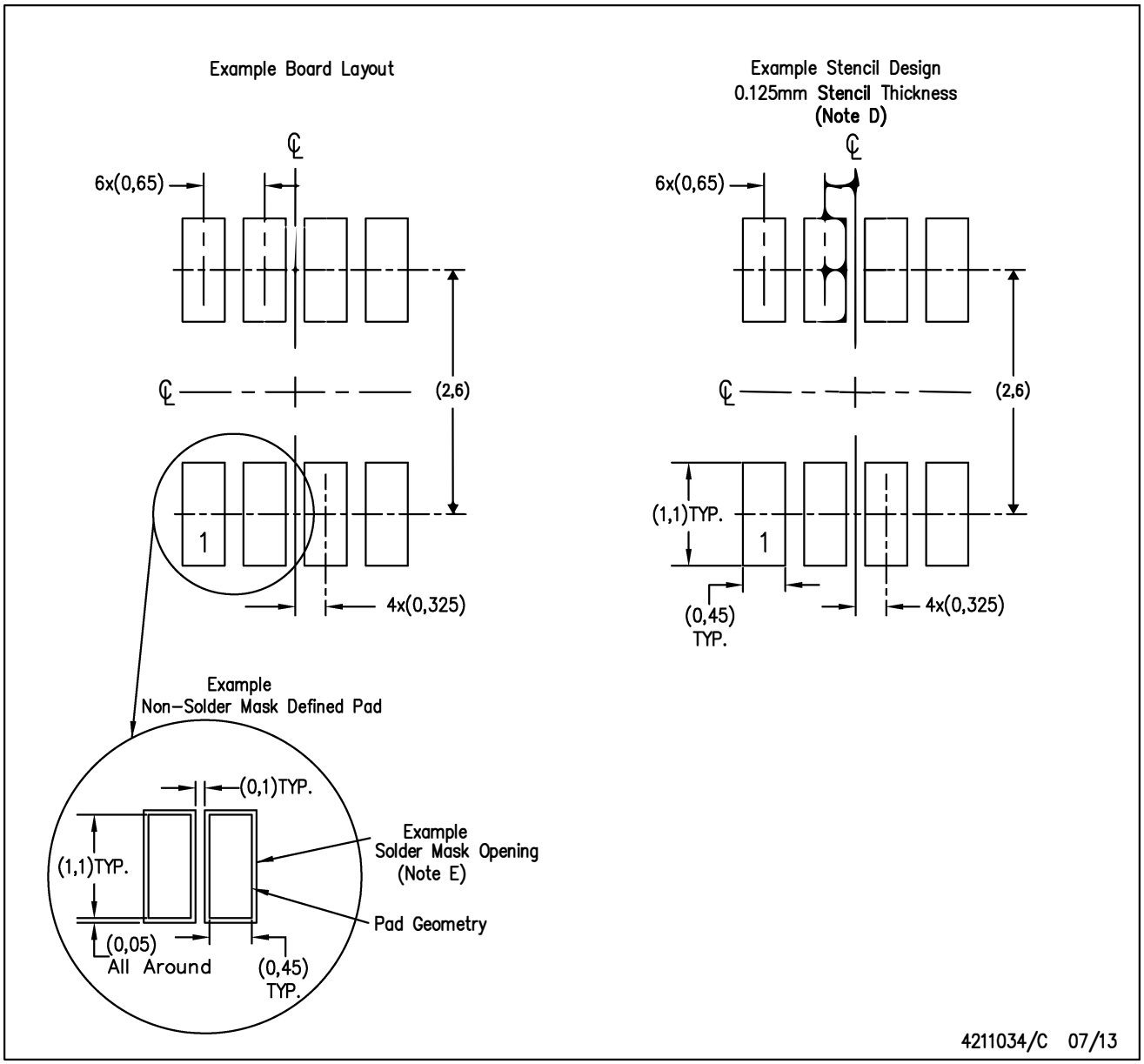
NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
D. Publication IPC–7351 is recommended for alternate designs.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC–7525 for other stencil recommendations.

DCN (R–PDSO–G8) PLASTIC SMALL–OUTLINE PACKAGE (DIE DOWN)



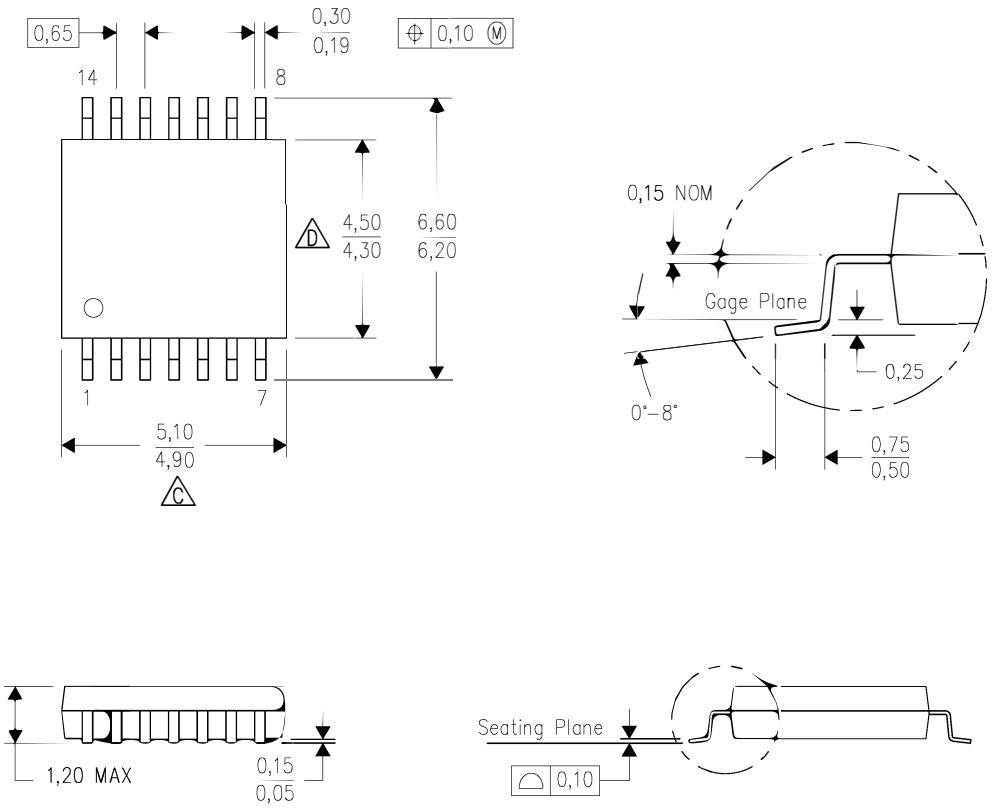
NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Package outline exclusive of metal burr & dambar protrusion/intrusion.
D. Package outline inclusive of solder plating.
E. A visual index feature must be located within the Pin 1 index area.
F. Falls within JEDEC MO–178 Variation BA.
G. Body dimensions do not include flash or protrusion. Mold flash and protrusion shall not exceed 0.25 per side.

DCN (R–PDSO–G8) PLASTIC SMALL–OUTLINE PACKAGE (DIE DOWN)



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC–7525.
E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

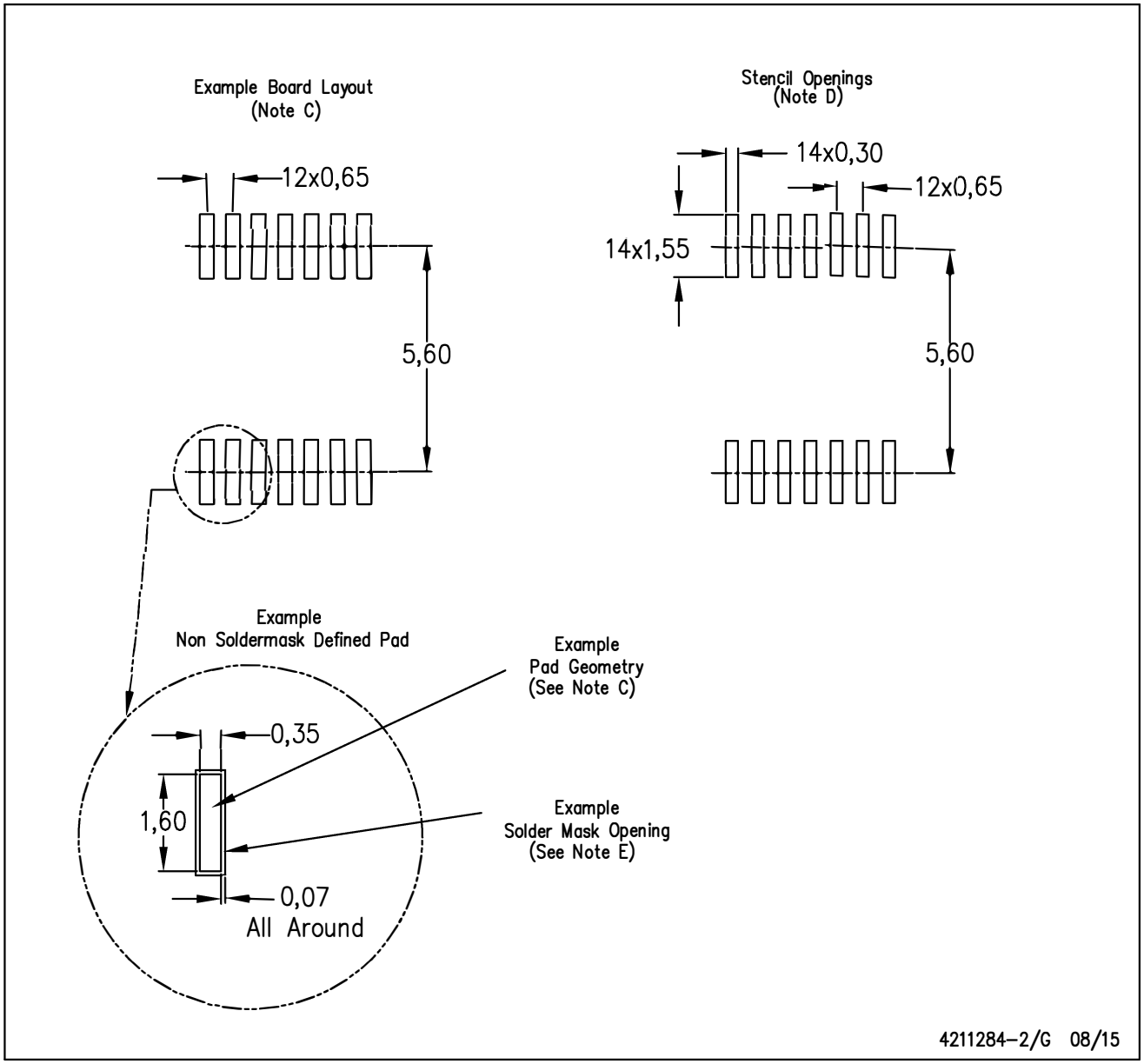
PW (R–PDSO–G14) PLASTIC SMALL OUTLINE



4040064–3/G 02/11

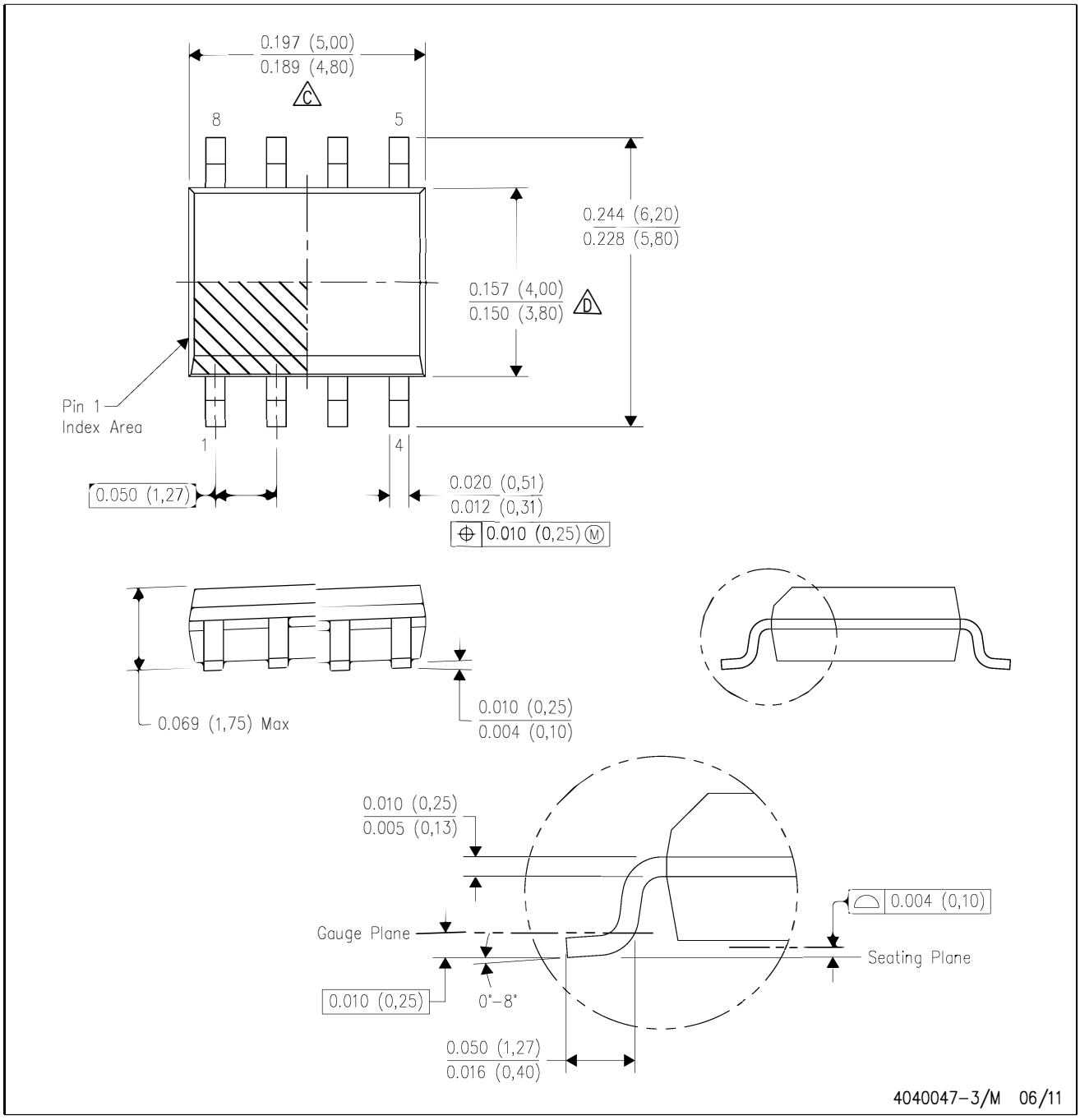
NOTES:
A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M–1994.
B. This drawing is subject to change without notice.
C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
E. Falls within JEDEC MO–153

PW (R–PDSO–G14) PLASTIC SMALL OUTLINE



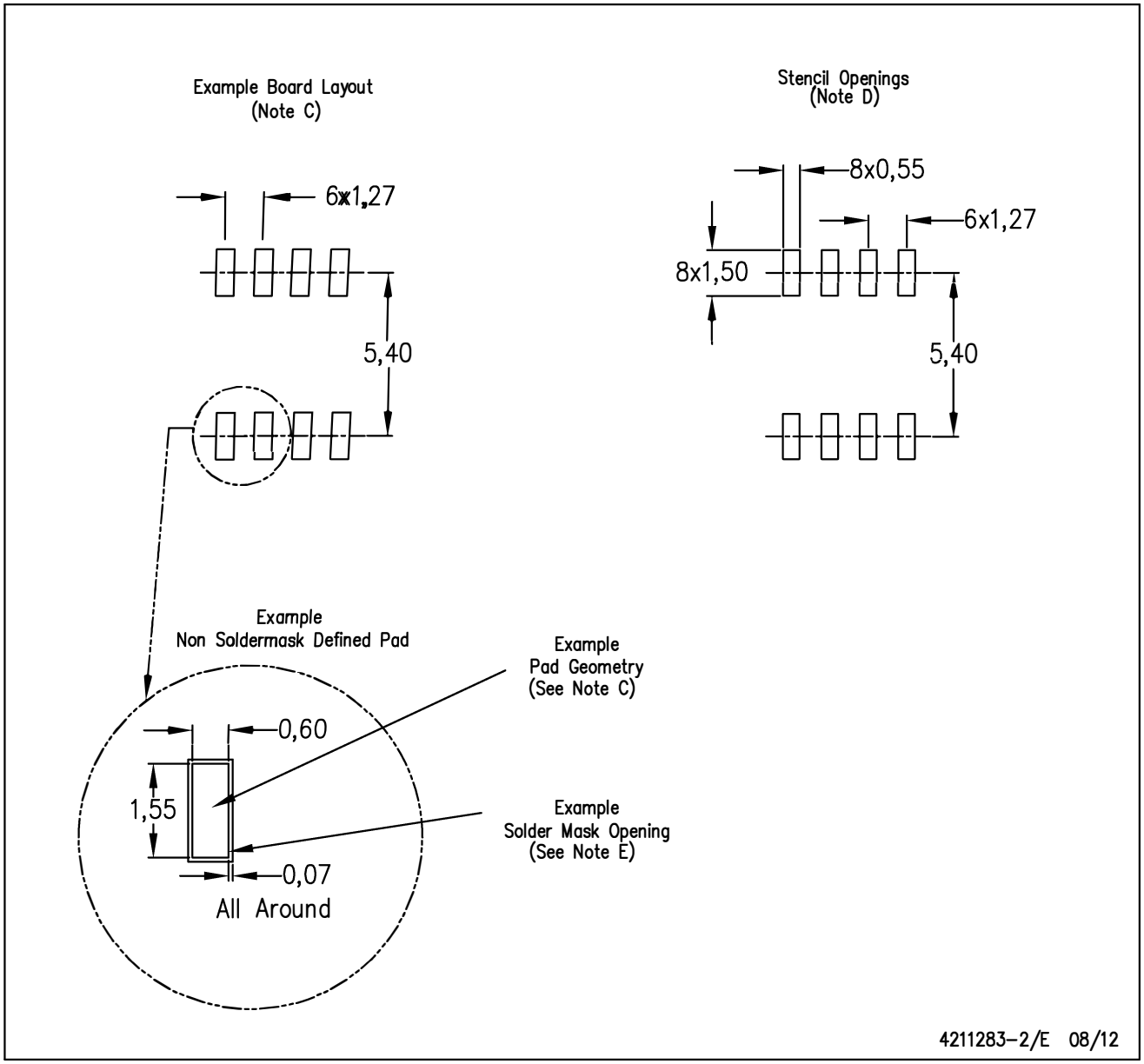
NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC–7525 for other stencil recommendations.
E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

D (R–PDSO–G8) PLASTIC SMALL OUTLINE



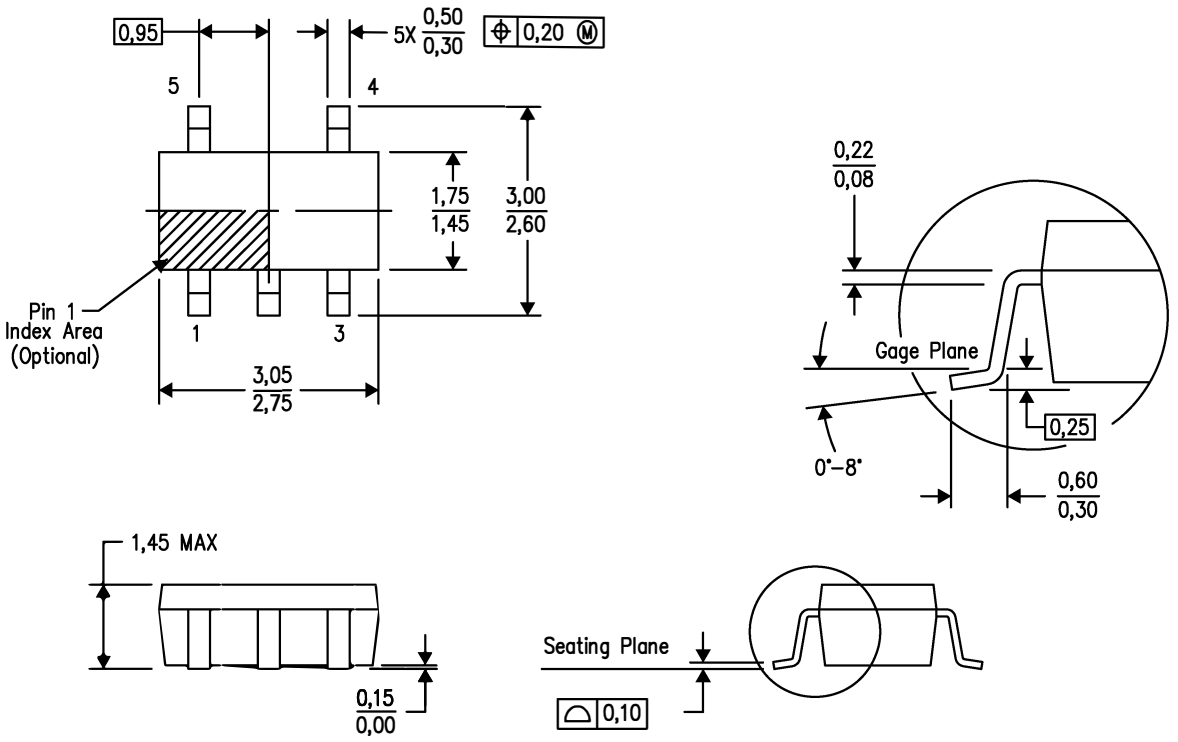
NOTES:
A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
D. Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
E. Reference JEDEC MS–012 variation AA.

D (R–PDSO–G8) PLASTIC SMALL OUTLINE



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC–7525 for other stencil recommendations.
E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

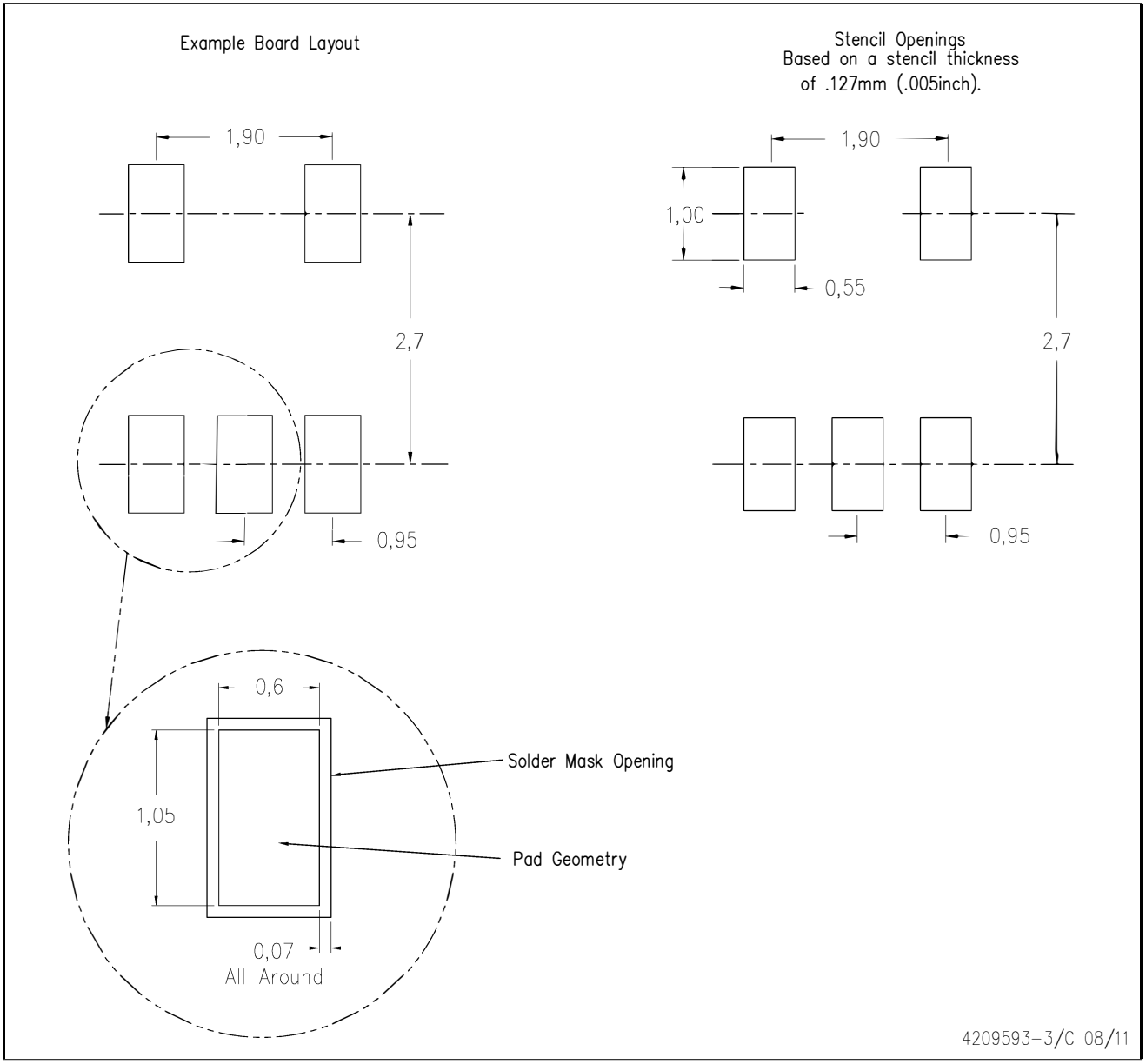
DBV (R–PDSO–G5) PLASTIC SMALL–OUTLINE PACKAGE



4073253–4/N 12/14

NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
D. Falls within JEDEC MO–178 Variation AA.

DBV (R–PDSO–G5) PLASTIC SMALL OUTLINE



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
D. Publication IPC–7351 is recommended for alternate designs.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC–7525 for other stencil recommendations.