

# SGM3209 High Voltage, Charge Pump DC/DC Converter

## **GENERAL DESCRIPTION**

The SGM3209 is a charge pump voltage converter. It converts a 3V to 18V input to a corresponding -3V to -18V output using the combination of few external components, eliminating inductors and their associated cost, size and EMI. In addition to a wider power supply input range, the SGM3209 can source output currents as high as 100mA. The switching frequency is resistor programmable from 120kHz to 1.25MHz.

The SGM3209 is recommended for designs requiring greater output current and/or lower input/output voltage drop. The SGM3209 enters into shutdown status by external enable control signal to reduce system power dissipation.

The SGM3209 is available in Green TDFN-2×2-8L and SOIC-8 packages. It operates over an ambient temperature range of -40°C to +85°C.

# FEATURES

- Wide Operating Range: 3V to 18V
- Output Current: 100mA
- 600kΩ Pull-Low Resistor on EN Pin
- 120kHz to 1.25MHz Programmable Oscillator Frequency
- No External Diodes Required
- Low Output Impedance: 15Ω (TYP) at I<sub>OUT</sub> = 20mA
- CMOS Construction
- -40°C to +85°C Operating Temperature Range
- Available in Green TDFN-2×2-8L and SOIC-8 Packages

### **APPLICATIONS**

Laptop Computers Disk Drives Process Instrumentation



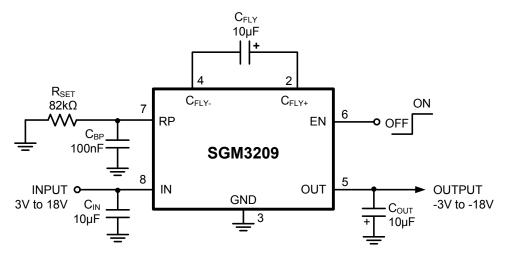


Figure 1. Typical Application Circuit



# PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION			PACKAGE MARKING	PACKING OPTION
SGM3209	SOIC-8	-40°C to +85°C	SGM3209YS8G/TR	SGM 3209YS8 XXXXX	Tape and Reel, 2500
	TDFN-2×2-8L	-40°C to +85°C	SGM3209YTDE8G/TR	3209 XXXX	Tape and Reel, 3000

NOTE: XXXX = Date Code. XXXXX = Date Code and Vendor Code.

Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

#### **ABSOLUTE MAXIMUM RATINGS**

IN to GND	0.3V to 22V
OUT to GND	(V <sub>IN</sub> + 0.3V) to 0.3V
EN, RP to GND	0.3V to 6V
Junction Temperature	150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (Soldering, 10sec) .	260°C
ESD Susceptibility	
HBM	8000V
MM	400V
CDM	1000V

### **RECOMMENDED OPERATING CONDITIONS**

Input Voltage Range	3V to 18V
Operating Temperature Range	40°C to +85°C

### **OVERSTRESS CAUTION**

Stresses beyond those listed may cause permanent damage to the device. Functional operation of the device at these or any other conditions beyond those indicated in the operational section of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

### ESD SENSITIVITY CAUTION

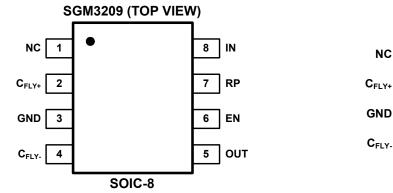
This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

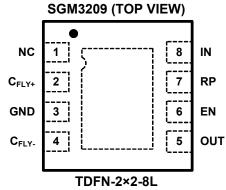
### DISCLAIMER

SG Micro Corp reserves the right to make any change in circuit design, specification or other related things if necessary without notice at any time.



## **PIN CONFIGURATIONS**





### **PIN DESCRIPTION**

P	N	NAME	FUNCTION
TDFN-2×2-8L	SOIC-8		FUNCTION
1	1	NC	No Connection.
2	2	C <sub>FLY+</sub>	Charge Pump Capacitor Positive Terminal.
3	3	GND	Ground Terminal.
4	4	C <sub>FLY-</sub>	Charge Pump Capacitor Negative Terminal.
5	5	OUT	Output Voltage.
6	6	EN	Enable Control Pin. Logic HIGH to enable chip and logic LOW to shut down chip.
7	7	RP	Switching Frequency Setting. Connect one resistor ( $R_{SET}$ ) between RP pin and GND to program switching frequency. The resistor should be located very close to this pin. Frequency range is from 120kHz to 1.25MHz.
8	8	IN	Supply Input.
Exposed Pad	_	NC	No Connection.

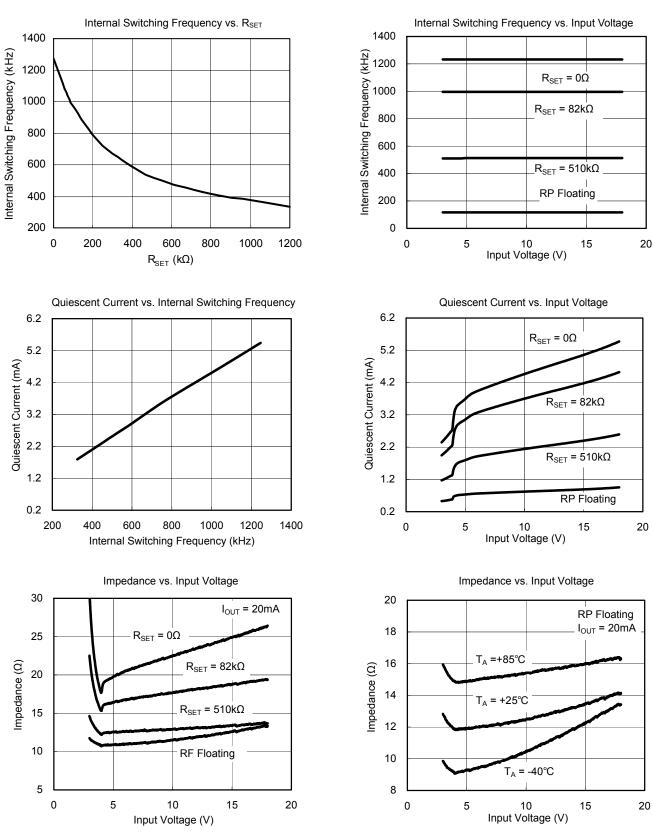
# **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = 15V, C_{IN} = C_{FLY} = C_{OUT} = 10\mu$ F, Full = -40°C to +85°C. Typical values are at T<sub>A</sub> = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	TEMP	MIN	TYP	MAX	UNITS
Input Voltage Range	V <sub>IN</sub>		Full	3		18	V
Maximum Output Current Range at OUT Pin	I <sub>OUT</sub>		+25°C	100			mA
Output Voltage	Vout		+25°C		-V <sub>IN</sub>		V
		I <sub>OUT</sub> = 100mA, RP Floating	+25°C		250		
Output Voltage Ripple	V <sub>PP</sub>	$I_{OUT}$ = 100mA, $R_{SET}$ = 82k $\Omega$	+25°C		40		$mV_{P-P}$
		I <sub>OUT</sub> = 100mA, R <sub>SET</sub> = 0Ω	+25°C		35		
		RP Floating, I <sub>OUT</sub> = 0mA	+25°C		0.9	1.4	
Quiescent Current		$R_{SET}$ = 82k $\Omega$ , $I_{OUT}$ = 0mA	+25°C		4.3	5.5	mA
	Ιq	$R_{SET} = 0\Omega, I_{OUT} = 0mA$	+25°C		5.3	7.0	
		Shutdown Mode, V <sub>EN</sub> = 0V	+25°C		0.5	1.2	μA
		RP Floating	+25°C	100	120	145	kHz
Internal Switching Frequency	f <sub>osc</sub>	R <sub>SET</sub> = 82kΩ	+25°C	850	1000	1160	
		R <sub>SET</sub> = 0Ω	+25°C	1050	1250	1520	
		I <sub>OUT</sub> = 20mA, RP Floating	+25°C		15	19	
Impedance	Ro	$I_{OUT}$ = 20mA, $R_{SET}$ = 82k $\Omega$	+25°C		23	38	
		$I_{OUT}$ = 20mA, $R_{SET}$ = 0 $\Omega$	+25°C		33	57	
EN Logic Level High	V <sub>IH</sub>	$V_{IN} = 3V$ to $18V$	Full	1.4			V
EN Logic Level Low	VIL	$V_{IN} = 3V$ to $18V$	Full			0.6	V
EN Internal Pull Low Resistance	R <sub>PULL-LOW</sub>		+25°C		600		kΩ
Thermal Shutdown Temperature	T <sub>SHDN</sub>		•		150		°C
Thermal Shutdown Hysteresis	$\Delta T_{SHDN}$				20		°C

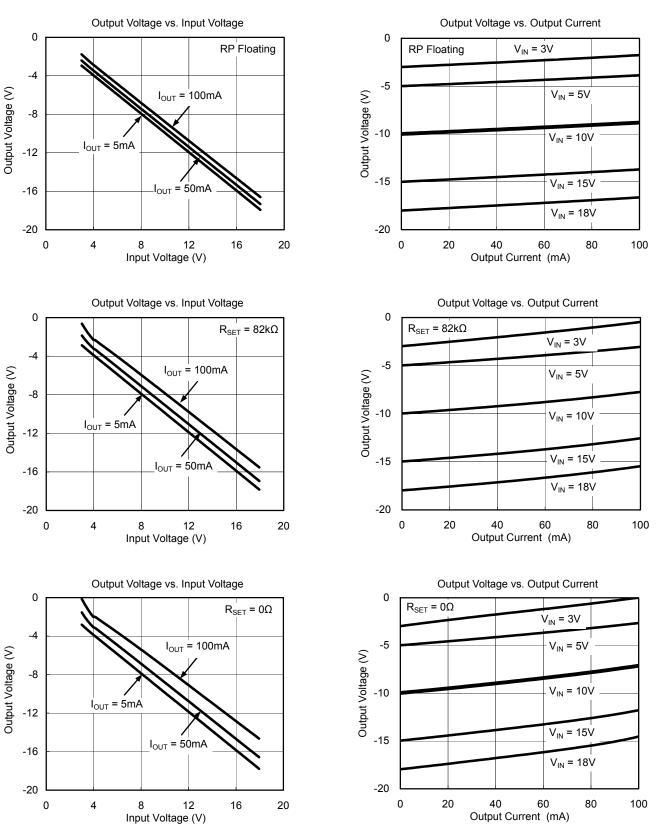


 $V_{IN}$  = 15V,  $C_{IN}$  =  $C_{FLY}$  =  $C_{OUT}$  = 10µF,  $T_A$  = +25°C, unless otherwise noted.



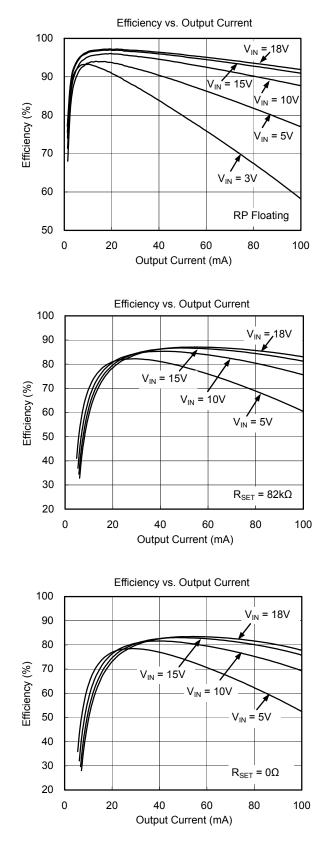
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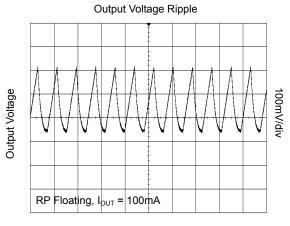
 $V_{IN}$  = 15V,  $C_{IN}$  =  $C_{FLY}$  =  $C_{OUT}$  = 10µF,  $T_A$  = +25°C, unless otherwise noted.



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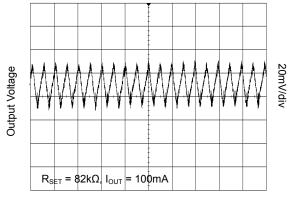
 $V_{\text{IN}}$  = 15V,  $C_{\text{IN}}$  =  $C_{\text{FLY}}$  =  $C_{\text{OUT}}$  = 10 $\mu\text{F},$   $T_{\text{A}}$  = +25°C, unless otherwise noted.



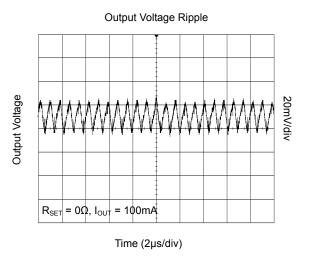


Time (10µs/div)

Output Voltage Ripple

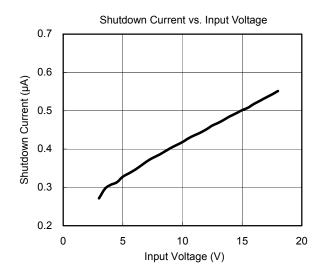


Time (2µs/div)



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 $V_{\text{IN}}$  = 15V,  $C_{\text{IN}}$  =  $C_{\text{FLY}}$  =  $C_{\text{OUT}}$  = 10 $\mu\text{F},$   $T_{\text{A}}$  = +25°C, unless otherwise noted.





### **DETAILED DESCRIPTION**

#### **Operating Principle**

The SGM3209 charge pump inverts the voltage applied to the input. For the best performance, use low equivalent series resistance (ESR) capacitors (e.g., ceramic). During the first half-cycle, switches S2 and S4 open, switches S1 and S3 close, and capacitor  $C_{FLY}$  charges to the voltage at  $V_{IN}$ . During the second half-cycle, S1 and S3 open and S2 and S4 close. This connects the positive terminal of  $C_{FLY}$  to GND and the negative to  $V_{OUT}$ . By connecting  $C_{FLY}$  in parallel,  $C_{OUT}$  is charged negative. The actual voltage at the output is more positive than  $-V_{IN}$ , since switches S1 - S4 have resistance and the load drains charge from  $C_{OUT}$ .

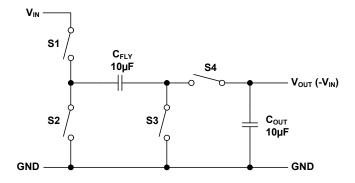


Figure 2. Operating Principle

#### Charge Pump Output Resistance

The SGM3209 device is not voltage regulator. The charge pump output source resistance is approximately 15 $\Omega$  at room temperature (with V<sub>IN</sub> = 15V), and V<sub>OUT</sub> approaches -15V when lightly loaded. V<sub>OUT</sub> will droop toward GND as load current increases.

$$V_{OUT} = -(V_{IN} - R_{OUT} \times I_{OUT})$$

$$R_{\text{OUT}} \approx \frac{1}{f_{\text{OSC}} \times C_{\text{FLY}}} + 4 \left( 2R_{\text{SWITCH}} + \text{ESR}_{\text{CFLY}} \right) + \text{ESR}_{\text{COUT}}$$

Where:

 $R_{OUT}$  = output resistance of the converter  $R_{SWITCH}$  = resistance of a single MOSFET-switch inside the converter f = oscillator frequency

 $f_{OSC}$  = oscillator frequency.

#### **Efficiency Considerations**

The power efficiency of a switched-capacitor voltage converter is affected by three factors: the internal losses in the converter IC, the resistive losses of the capacitors, and the conversion losses during charge transfer between the capacitors. The internal losses are associated with the IC's internal functions, such as driving the switches, oscillator, etc. These losses are affected by operating conditions such as input voltage, temperature, and frequency. The next two losses are associated with the voltage converter circuit's output resistance. Switch losses occur because of the on-resistance of the MOSFET switches in the IC. Charge pump capacitor losses occur because of their ESR. The relationship between these losses and the output resistance is as follows:

$$P_{CAPACITOR LOSSES} + P_{CONVERSION LOSSES} = I_{OUT}^2 \times R_{OUT}$$

The first term is the effective resistance from an ideal switched-capacitor circuit. Conversion losses occur during the charge transfer between  $C_{FLY}$  and  $C_{OUT}$  when there is a voltage difference between them. The power loss is:

P<sub>CONVERSION LOSS</sub> =

$$\left\lfloor \frac{1}{2} \times C_{FLY} \left( V_{IN}^{2} - V_{OUT}^{2} \right) + \frac{1}{2} C_{OUT} \left( V_{PP}^{2} - 2V_{OUT} V_{PP} \right) \right\rfloor \times f_{OSC}$$

The efficiency of the SGM3209 is dominated by their quiescent supply current at low output current and by their output impedance at higher current.

$$\eta \cong \frac{I_{\text{OUT}}}{I_{\text{OUT}} + I_{\text{Q}}} \left( 1 - \frac{I_{\text{OUT}} \times R_{\text{OUT}}}{V_{\text{IN}}} \right)$$

Where:  $I_Q$  = quiescent current.



### **DETAILED DESCRIPTION**

#### **Capacitor Selection**

To maintain the lowest output resistance, use capacitors with low ESR (see Table 1). The charge pump output resistance is a function of  $C_{FLY}$ 's and  $C_{OUT}$ 's ESR. Therefore, minimizing the charge pump capacitor's ESR minimizes the total output resistance. The capacitor values are closely linked to the required output current and the output noise and ripple requirements. It is possible to only use 10µF capacitors of the same type.

#### **Table 1. Recommended Capacitor Values**

V <sub>IN</sub> (V)	I <sub>OUT</sub> (mA)	C <sub>IN</sub> (μF)	C <sub>FLY</sub> (μF)	С <sub>оит</sub> (µF)
3 to 18	100	10	10	10

#### Input Capacitor (C<sub>IN</sub>)

Bypass the incoming supply to reduce its AC impedance and the impact of the SGM3209 switching noise. The recommended bypassing depends on the circuit configuration and where the load is connected. When the inverter is loaded from OUT to GND, current from the supply switches between  $2 \times I_{OUT}$  and zero. Therefore, use a large bypass capacitor (e.g., equal to the value of  $C_{FLY}$ ) if the supply has high AC impedance. When the inverter is loaded from IN to OUT, the circuit draws  $2 \times I_{OUT}$  constantly, except for short switching spikes. A  $0.1\mu$ F bypass capacitor is sufficient.

#### Flying Capacitor (C<sub>FLY</sub>)

Increasing the flying capacitor's size reduces the output resistance. Small values increase the output resistance. Above a certain point, increasing  $C_{FLY}$ 's capacitance has a negligible effect, because the output resistance becomes dominated by the internal switch resistance and capacitor ESR.

#### **Output Capacitor (COUT)**

Increasing the output capacitor's size reduces the output ripple voltage. Decreasing its ESR reduces both output resistance and ripple. Smaller capacitance values can be used with light loads if higher output ripple can be tolerated. Use the following equation to calculate the peak-to-peak ripple.

$$V_{PP} = \frac{I_{OUT}}{f_{OSC} \times C_{OUT}} + 2 \times I_{OUT} \times ESR_{COUT}$$

### **Operating Frequency**

The operating frequency of the SGM3209 is determined by an external resistor that is connected between the RP pin and ground. The value of the resistor sets the ramp current that is used to charge and discharge an internal timing capacitor within the oscillator. The  $R_{SET}$  resistor value can be determined by examining the Internal Switching Frequency vs.  $R_{SET}$  curve. To filtering switching noise, a 100nF capacitor should be connected between RP pin and GND.



## **APPLICATION INFORMATION**

#### Simple Negative Voltage Converter

The majority of applications will undoubtedly utilize the SGM3209 for generation of negative supply voltages. Figure 3 shows typical connections to provide a negative supply where a positive supply of +3V to +18V is available.

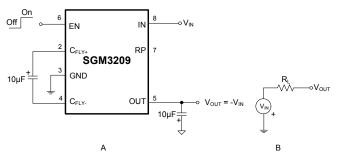


Figure 3. Simple Negative Converter and Its Output Equivalent

#### **Paralleling Devices**

Any number of the SGM3209 voltage converters may be paralleled to reduce output resistance (Figure 4). The reservoir capacitor,  $C_{OUT}$ , serves all devices, while each device requires its own pump capacitor,  $C_{FLY}$ . The resultant output resistance would be approximately:

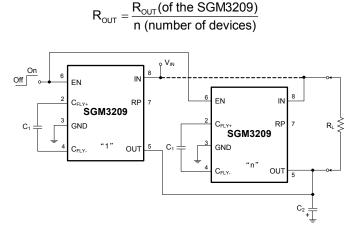


Figure 4. Paralleling Devices Lowers Output Impedance

#### **Cascading Devices**

The SGM3209 may be cascaded as shown (Figure 5) to produce larger negative multiplication of the initial supply voltage. However, due to the finite efficiency of each device, the practical limit is 10 devices for light loads. The output voltage is defined by:

$$V_{OUT}$$
 = -n ×  $V_{IN}$ 

where n is an integer representing the number of devices cascaded. The resulting output resistance would be approximately the weighted sum of the individual SGM3209  $R_{OUT}$  values.

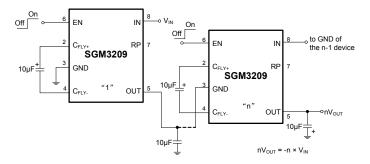


Figure 5. Increased Output Voltage by Cascading Devices

### **APPLICATION INFORMATION**

#### **Positive Voltage Doubling**

The SGM3209 may be employed to achieve positive voltage doubling using the circuit shown in Figure 6. In this application, the pump inverter switches of the SGM3209 are used to charge  $C_{OUT}$  to a voltage level of  $V_{IN} - V_F$  (where  $V_{IN}$  is the supply voltage and  $V_F$  is the forward voltage on  $C_{FLY}$  plus the supply voltage  $V_{IN}$  applied through diode  $D_2$  to capacitor  $C_{OUT}$ ). The voltage thus created on  $C_{OUT}$  becomes ( $2V_{IN}$ ) - ( $2V_F$ ), or twice the supply voltage minus the combined forward voltage drops of diodes  $D_1$  and  $D_2$ .

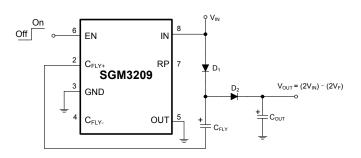


Figure 6. Positive Voltage Multiplier

#### Combined Negative Voltage Conversion and Positive Supply Multiplication

Figure 7 combines the functions shown in Figure 3 and Figure 6 to provide negative voltage conversion and positive voltage doubling simultaneously. This approach would be, for example, suitable for generating +9V and -5V from an existing +5V supply. In this instance, capacitors  $C_1$  and  $C_3$  perform the pump and reservoir functions, respectively, for the generation of the negative voltage, while capacitors C2 and C4 are pump and reservoir, respectively, for the doubled positive voltage. There is a penalty in this configuration which combines both functions, however, in that the source impedances of the generated supplies will be somewhat higher due to the finite impedance of the common charge pump driver at pin 2 of the device.

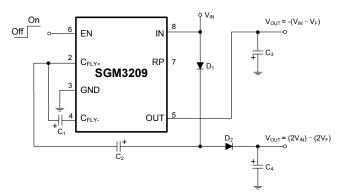
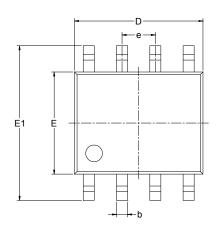
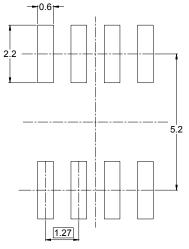


Figure 7. Combined Negative Converter and Positive Doubler

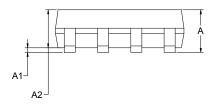
# PACKAGE OUTLINE DIMENSIONS

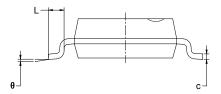
# SOIC-8





RECOMMENDED LAND PATTERN (Unit: mm)



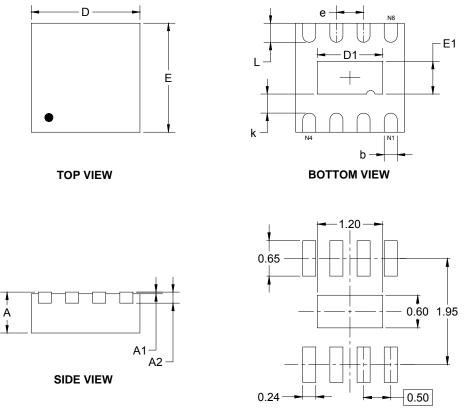


Symbol		nsions meters	Dimensions In Inches		
	MIN	MAX	MIN	MAX	
A	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
е	1.27 BSC		0.050	BSC	
L	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	



# PACKAGE OUTLINE DIMENSIONS

## TDFN-2×2-8L



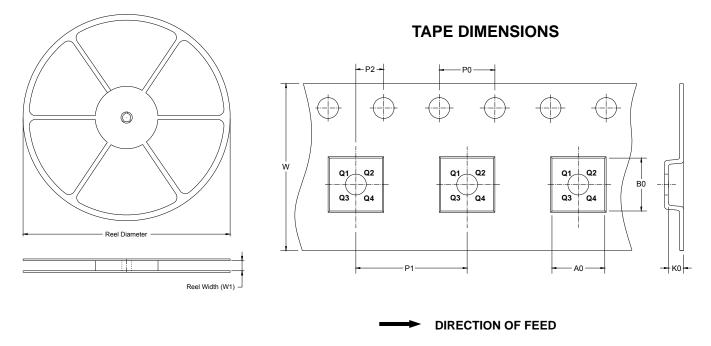
RECOMMENDED LAND PATTERN (Unit: mm)

Symbol		nsions meters	Dimensions In Inches		
	MIN	MAX	MIN	MAX	
A	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A2	0.203	B REF	0.008 REF		
D	1.900	2.100	0.075	0.083	
D1	1.100	1.300	0.043	0.051	
E	1.900	2.100	0.075	0.083	
E1	0.500	0.700	0.020	0.028	
k	0.200	) MIN	300.0	3 MIN	
b	0.180	0.300	0.007	0.012	
е	0.500	0.500 TYP		) TYP	
L	0.250	0.450	0.010	0.018	



## TAPE AND REEL INFORMATION

### **REEL DIMENSIONS**



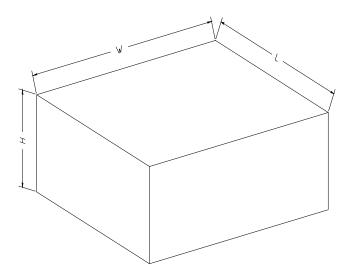
NOTE: The picture is only for reference. Please make the object as the standard.

### **KEY PARAMETER LIST OF TAPE AND REEL**

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOIC-8	13″	12.4	6.4	5.4	2.1	4.0	8.0	2.0	12.0	Q1
TDFN-2×2-8L	7″	9.5	2.30	2.30	1.10	4.00	4.00	2.00	8.00	Q1



### **CARTON BOX DIMENSIONS**



NOTE: The picture is only for reference. Please make the object as the standard.

### **KEY PARAMETER LIST OF CARTON BOX**

Reel Type	Length Width (mm) (mm)		Height (mm)	Pizza/Carton	
7" (Option)	368	227	224	8	
7″	442	410	224	18	
13″	386	280	370	5	DD0002

